

Bi-CMOS Linear Integrated Circuit Silicon Monolithic

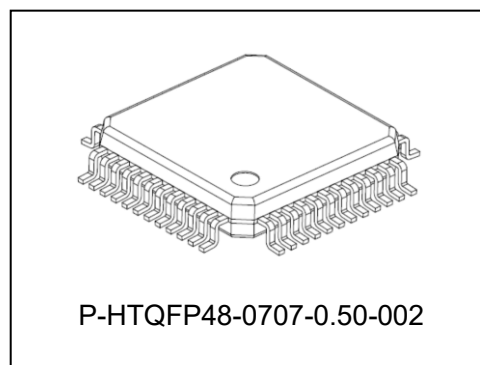
# TB9M030FG

IC for automotive three-phase brushless DC motor

## 1. Description

The TB9M030FG is an integrated device that combines a microcontroller unit (MCU) and a gate driver IC for automotive brushless DC (BLDC) motor applications, which incorporates Toshiba's unique Vector Engine (VE) for easy implementation of vector control while reducing CPU load.

The TB9M030FG is designed to be used with external N-channel FET pairs, making it suitable for BLDC motors with a wide range of output. The TB9M030FG also incorporates a CPU and a flash ROM, making it possible to program a control method and parameters according to the motor and application requirements. The TB9M030FG can be configured to transition to Standby mode in the idle state in order to reduce power consumption.



weight: 0.14 g (typ.)

## 2. Applications

For automotive (electric pumps and fans), three-phase brushless motor control and drive.

## 3. Features

- Integrated MCU and gate drivers enable downsizing of the system
- Built-in vector engine for our original sensorless control

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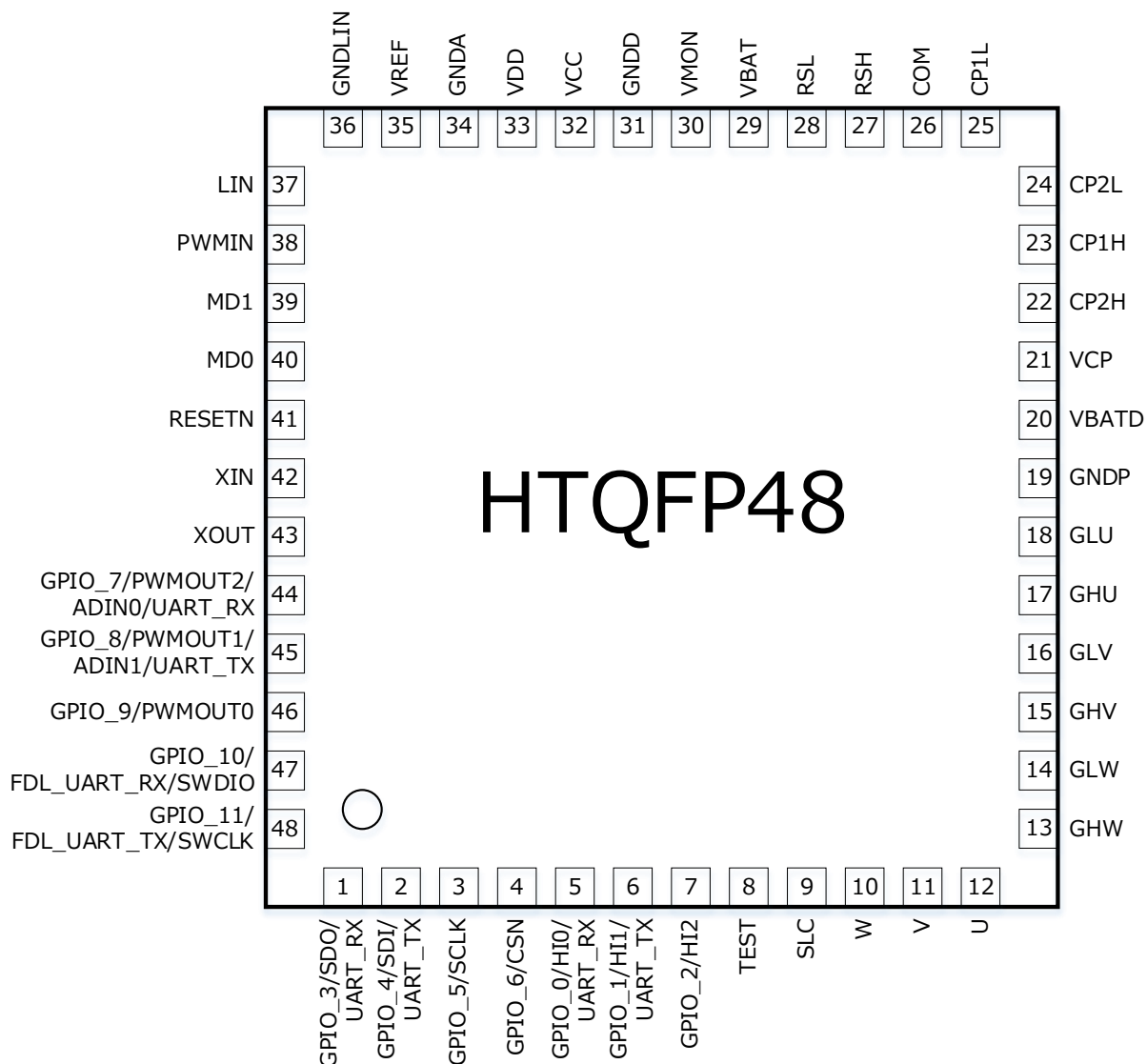
## 4. Functions

- 32 bit Arm® Cortex®-M0 processor
  - Serial Wire Debug Support
  - 32ch Interrupt Controller and 1 cycle multiplier
  - Up to 40 MHz clock frequency
- Single Bus Master System
- 12 KBytes ROM(BootLoader, Flash API) (incECC SEC/DED)
- 64 KBytes Code Flash(incl. ECC SEC/DED)
- 4 KBytes SRAM (incl. ECC SEC/DED)
- 32-bit Compare Timers (DTIMER)
- 28-bit Capture Timer (8 inputs, 6 measurement )
- Watchdog Timer(WATCHDOG)
- Power saving modes ( CPU Sleep, Standby )
- 4ch Pulse Width Modulator Generator(PWMGEN)
- 12 General-purpose I/O Ports (GPIO)
- 10-bit A/D Converter (GADC) with 2 analog inputs + internal temperature, VMON
- 12-bit A/D Converter (MADC)
- Vector Engine(VE)
- Programmable Motor Driver(PMD)
- Encoder(ENC)
- LDOs (LDO5V, LDO15V )
- Power On Reset ( POR5V, PORL )
- 2 on-chip OSCs (IOSCH, IOSCL)
- External OSC
- PLL
- LIN ISO17987/SAEJ2602 transceiver + controller
- 2 full duplex serial interfaces (UART1, UART2), UART1 with LIN support
- SPI-I/F
- MOSFET driver including charge pump
- High-speed operational amplifier for motor current sensing via shunt
- Over current protection (LDO, MOSFET driver)
- Over temperature protection
- Package HTQFP48-0707-0.50
- Single power supply from 6.0V to 18V
- Temperature Range Tj = -40°C to +175°C
- Green package (RoHS compliant)
- AEC-Q100 grade 0 qualified



## 6. Pin Assignments

(Top view)



**Figure 6.1 Pin Assignment Diagram**

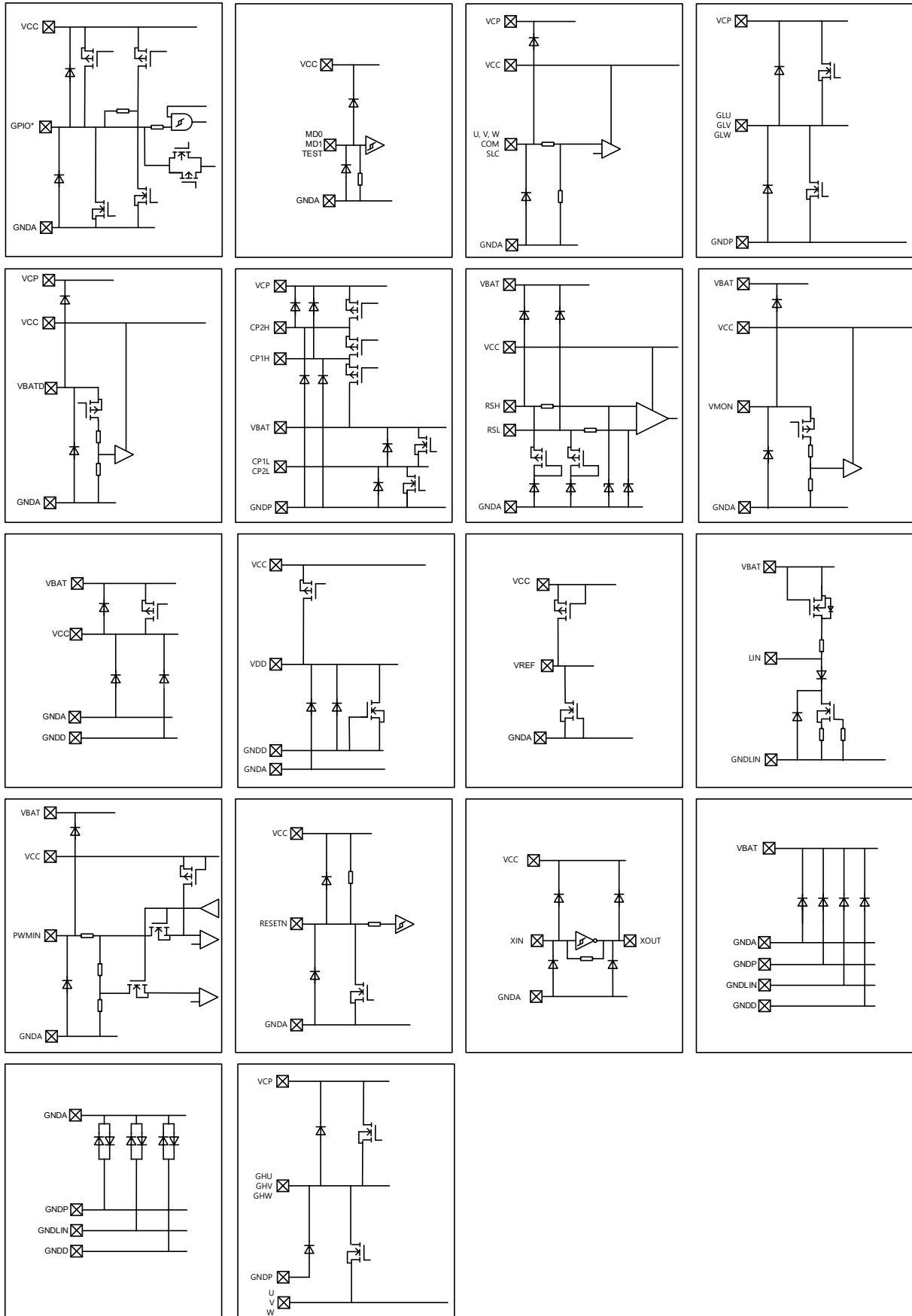
## 7. Pin Description

Table 7.1 Pin Description

Pin	Symbol	I/O	Description
1	GPIO_3/SDO/UART_RX	I/O	General-purpose I/O port / SPI data output/UART_RX
2	GPIO_4/SDI/UART_TX	I/O	General-purpose I/O port / SPI data input/UART_TX
3	GPIO_5/SCLK	I/O	General-purpose I/O port / SPI data clock
4	GPIO_6/CSN	I/O	General-purpose I/O port / SPI chip select
5	GPIO_0/HI0/UART_RX	I/O	General-purpose I/O port/ Hall sensor input/UART_RX
6	GPIO_1/HI1/UART_TX	I/O	General-purpose I/O port/ Hall sensor input/UART_TX
7	GPIO_2/HI2	I/O	General-purpose I/O port/ Hall sensor input
8	TEST	I	Test mode select
9	SLC	I	Low-side FET source input
10	W	I	Phase-W motor connection pin
11	V	I	Phase-V motor connection pin
12	U	I	Phase-U motor connection pin
13	GHW	O	Phase-W high-side FET gate output
14	GLW	O	Phase-W low-side FET gate output
15	GHV	O	Phase-V high-side FET gate output
16	GLV	O	Phase-V low-side FET gate output
17	GHU	O	Phase-U high-side FET gate output
18	GLU	O	Phase-U low-side FET gate output
19	GNDP	-	Power ground
20	VBATD	I	Battery voltage regulator input for driver circuit
21	VCP	O	Voltage regulator output for charge pump
22	CP2H	O	Boost capacitor connection pin for charge pump
23	CP1H	O	Boost capacitor connection pin for charge pump
24	CP2L	O	Boost capacitor connection pin for charge pump
25	CP1L	O	Boost capacitor connection pin for charge pump
26	COM	I	Motor pseudo-neutral point input
27	RSH	I	VCC side of current-sense resistor
28	RSL	I	Ground side of current-sense resistor
29	VBAT	I	Battery voltage regulator input
30	VMON	I	Battery voltage regulator input (for ADC input)
31	GNDD	-	Logic ground
32	VCC	O	5-volt regulator output
33	VDD	O	1.5-volt regulator output
34	GND A	-	Analog ground
35	VREF	I	ADC reference voltage input
36	GNDLIN	-	Ground for LIN Phy
37	LIN	I/O	LIN bus line
38	PWMIN	I	PWM speed input
39	MD1	I	Mode select

40	MD0	I	Mode select
41	RESETN	I/O	Reset I/O
42	XIN	I	Crystal or ceramic oscillator connection pin
43	XOUT	O	Crystal or ceramic oscillator connection pin
44	GPIO_7/PWMOUT2/ADIN0/UART_RX	I/O	General-purpose I/O port /PWM Output/ 10 bit ADC input /UART_RX /External interrupt input
45	GPIO_8/PWMOUT1/ADIN1/UART_TX	I/O	General-purpose I/O port /PWM Output/ 10 bit ADC input /UART_TX
46	GPIO_9/PWMOUT0	I/O	General-purpose I/O port /PWM output / PWM Communication Output /External interrupt input
47	GPIO_10/FDL_UART_RX/SWDIO	I/O	General-purpose I/O port /UART RX for FDL_UART /SWDIO for Debug
48	GPIO_11/FDL_UART_TX/SWCLK	I/O	General-purpose I/O port / UART TX for FDL_UART/ SWCLK for Debug
-	EP	-	Exposed pad, connect to GND

**8. I/O Equivalent Circuits**

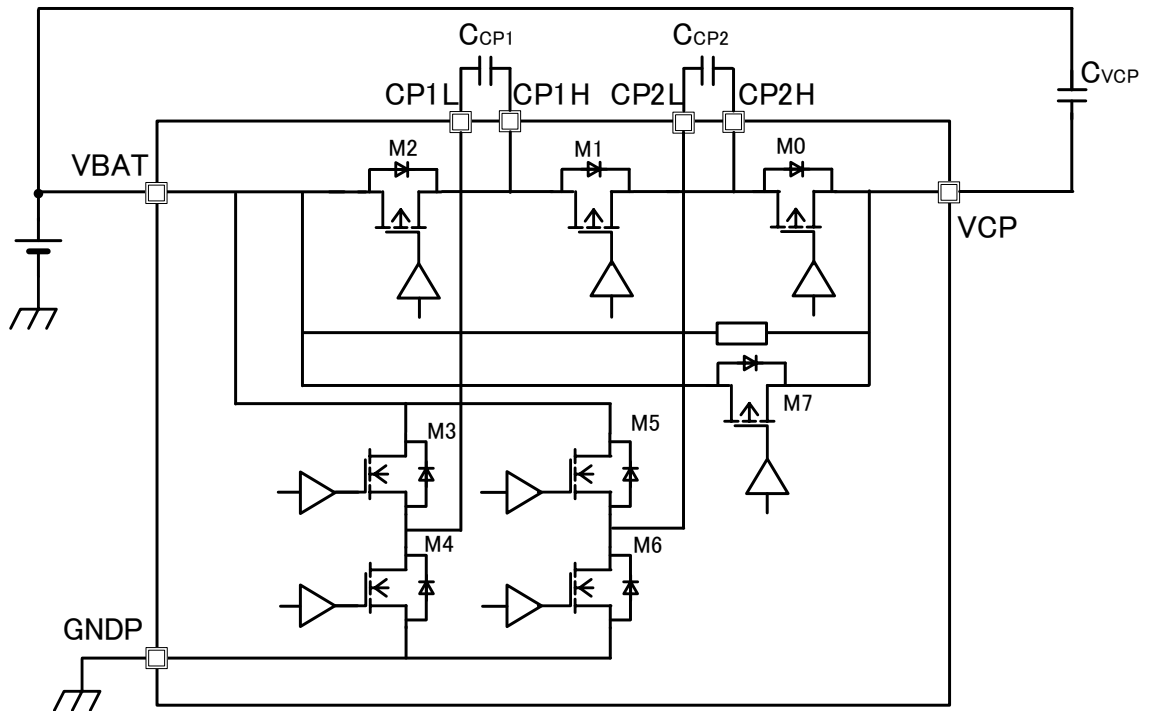


**Figure 8.1 I/O Equivalent Circuit Diagrams**



**9.2. Charge Pump Circuit**

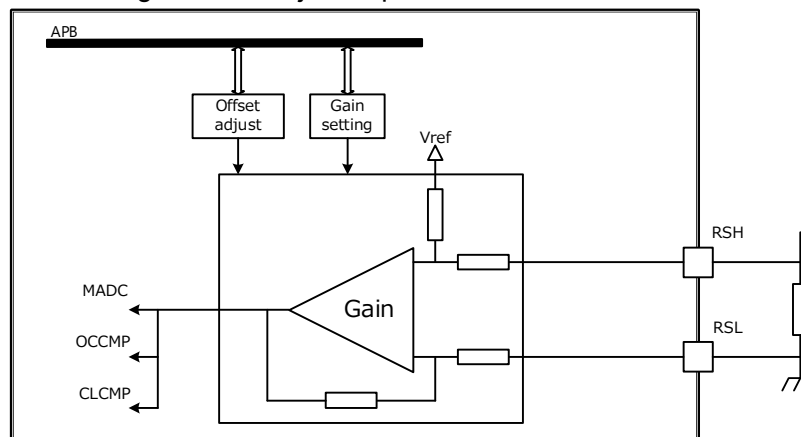
The CGP generates a voltage for the PREDRV to drive external high-side N-channel FETs.



**Figure 9.2.1 Charge Pump Circuit Diagram**

**9.3. Current-Sense Amplifier (CSAMP)**

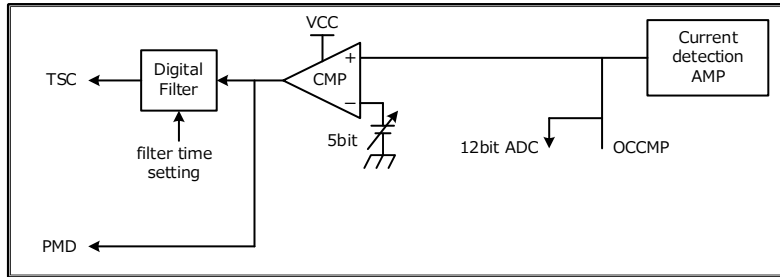
- The Current-Sense Amplifier (CSAMP) amplifies the voltage across an external shunt resistor by the gain programmed via a 3-bit register field.
- The amplified voltage is applied to the MADC, Overcurrent Detection Comparator (OCCMP), and Current Clamp Comparator (CLCMP).
- The output offset voltage can be adjusted prior.



**Figure 9.3.1 Current-Sense Amplifier Circuit Diagram**

**9.4. Current Clamp Comparator (CLCMP)**

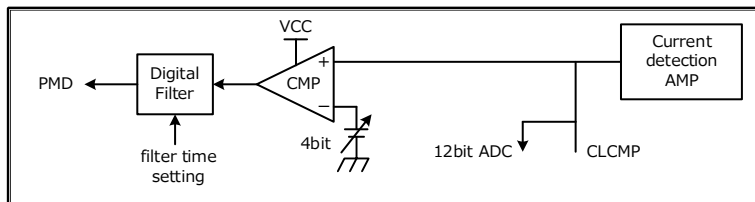
- The CLCMP limits current if the output from the Current-Sense Amplifier (CSAMP) exceeds the programmed voltage threshold.
- The output voltage of the CSAMP is applied to the CLCMP.
- This threshold is programmable in 32 steps (i.e., with a 5bit value)
- The CLCMP incorporates a digital filter to prevent malfunction caused by noise.



**Figure 9.4.1 Current Clamp Comparator Circuit Diagram**

**9.5. Overcurrent Detection Comparator (OCCMP)**

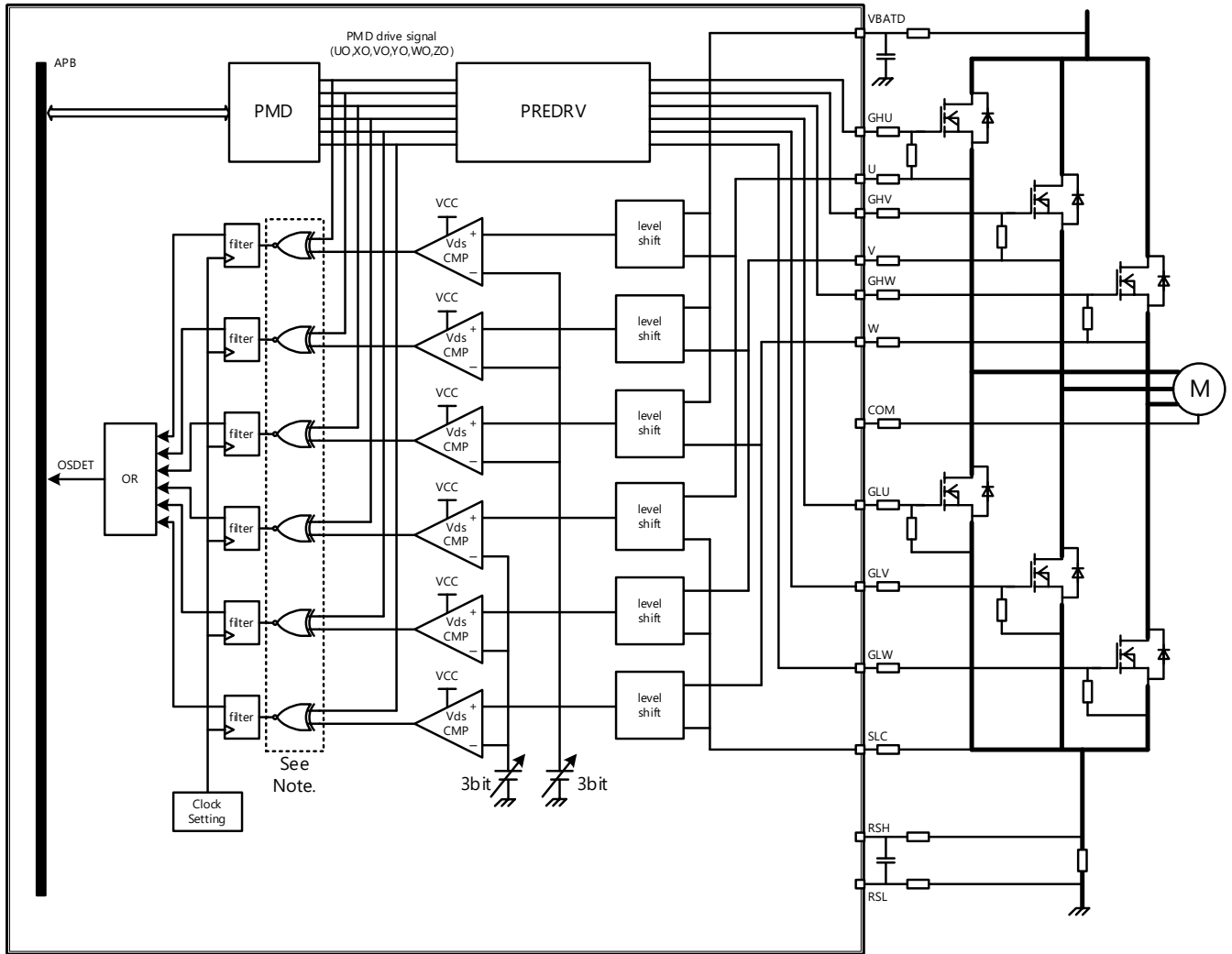
- The OCCMP reports an error when the output from the Current-Sense Amplifier (CSAMP) exceeds the programmed threshold.
- The OCCMP can be enabled and disabled via the ANASTBY register.
- The overcurrent threshold is programmable with a four-bit value via a special function register.
- The OCCMP incorporates a digital filter to prevent malfunction caused by noise.



**Figure 9.5.1 Overcurrent Detection Comparator Circuit Diagram**

**9.6. Open/Short Comparator (OSCMP) for External FETs**

The Open/Short Comparator detects open- and short-circuits involving external FETs.



**Figure 9.6.1 OSCMP Block Diagram**

**9.7. Vector Engine(VE)**

**Table 9.1 VE Function List**

Category	Function	Description
<b>Calculation</b>	Basic functions	Fixed-point calculation Vector control tasks Tasks for interfacing with the PMD and MADC
	Current control tasks	PI control of d-axis and q-axis currents <ul style="list-style-type: none"> <li>• Non-interference control</li> <li>• Output limiting based on scalar voltage values</li> </ul>
	SIN/COS Calculation 1 task	Calculates the sine and cosine values of phase $\theta$ <ul style="list-style-type: none"> <li>• Allows phase interpolation and clipping</li> </ul>
	SIN/COS Calculation 2 task	Calculates the sine and cosine values of phase $\theta$
	SIN/COS Calculation 3 task	Calculates the sine and cosine values of phase $\theta$ <ul style="list-style-type: none"> <li>• Allows phase interpolation and clipping</li> </ul>
	Output Voltage Transformation task	<ul style="list-style-type: none"> <li>• Coordinate axis transformation (inverse Park transformation)</li> </ul> 2 types of phase transformation (space vector modulation and inverse Clark transformation)
	Output Control task	Converts three-phase voltage into PWM output settings for the PMD (2 types) <ul style="list-style-type: none"> <li>• Allows output limiting</li> <li>• Allows dead-time compensation</li> </ul>
	Trigger Generation task	Calculates the PMD's AD conversion sampling timing from three-phase duty cycles (2 types)
	Current Correction Preparation task	Corrects detected current values for low inductance motor. Used in combination with Input Process 6 tasks.
	Input Process tasks	Reads conversion results from the MADC and converts them into fixed-point values (4 types) Supports current polarity determination (hysteresis/reverse hysteresis)
	Input current transformation task	<ul style="list-style-type: none"> <li>• Phase transformation (Clark transformation)</li> <li>• Coordinate axis transformation</li> </ul>
	Individual functions task	<ul style="list-style-type: none"> <li>• Arctangent (ATAN) calculation</li> <li>• Square-root calculation</li> <li>• No Operation(NOP)</li> </ul>
	Position Estimation task	<ul style="list-style-type: none"> <li>• Calculates electrical angular velocity and phase <math>\theta</math> from motor parameters, voltage, and current (2 types)</li> </ul>
	Position Sensor Input Process task	Calculates phase $\theta$ and electrical angular velocity from the inputs from a position sensor such as an encoder that generates multiple pulses per revolution
<b>Schedule management</b>	Program Schedule control	<ul style="list-style-type: none"> <li>• Program schedule capable of defining the order in which to execute tasks and their start-up control</li> <li>• Supports up to 32 tasks</li> </ul>
	Start-up control	<ul style="list-style-type: none"> <li>• Repetition start</li> </ul> Starts the input schedule upon completion of an AD conversion Starts executing an Input Process task upon ADC-complete interrupt when the VE is in the Standby state following the completion of an output schedule
<b>Interrupt control</b>	Schedule-complete interrupt	Generates an interrupt when a task with the END flag set to "1" is executed repeatedly for the specified number of times ([REPTIME]).
	Error interrupt	Interrupt generated when the VE receives a PWM interrupt from the PMD during the execution of an output schedule, judging it as an error.

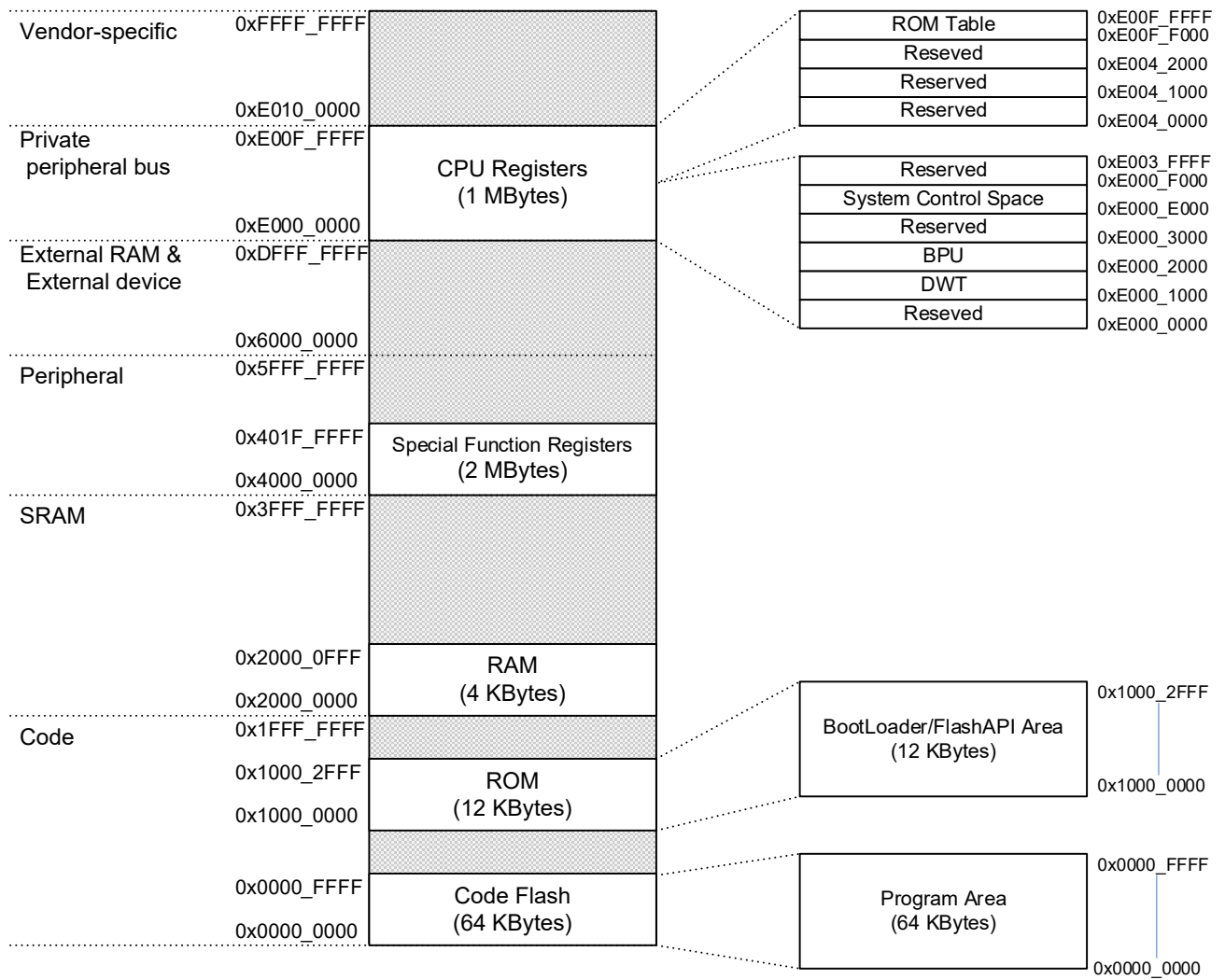
<b>Other</b>	Outputs for debugging	Outputs a signal that indicates that the task is running by toggling the debug output when the task starts and stops. For example, depending on the timing of the PMD debug output, the period indicating the operation may appear to be inverted (L output). This can be monitored by the debug output function of the PMD.
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## 9.8. Programmable Motor Driver(PMD)

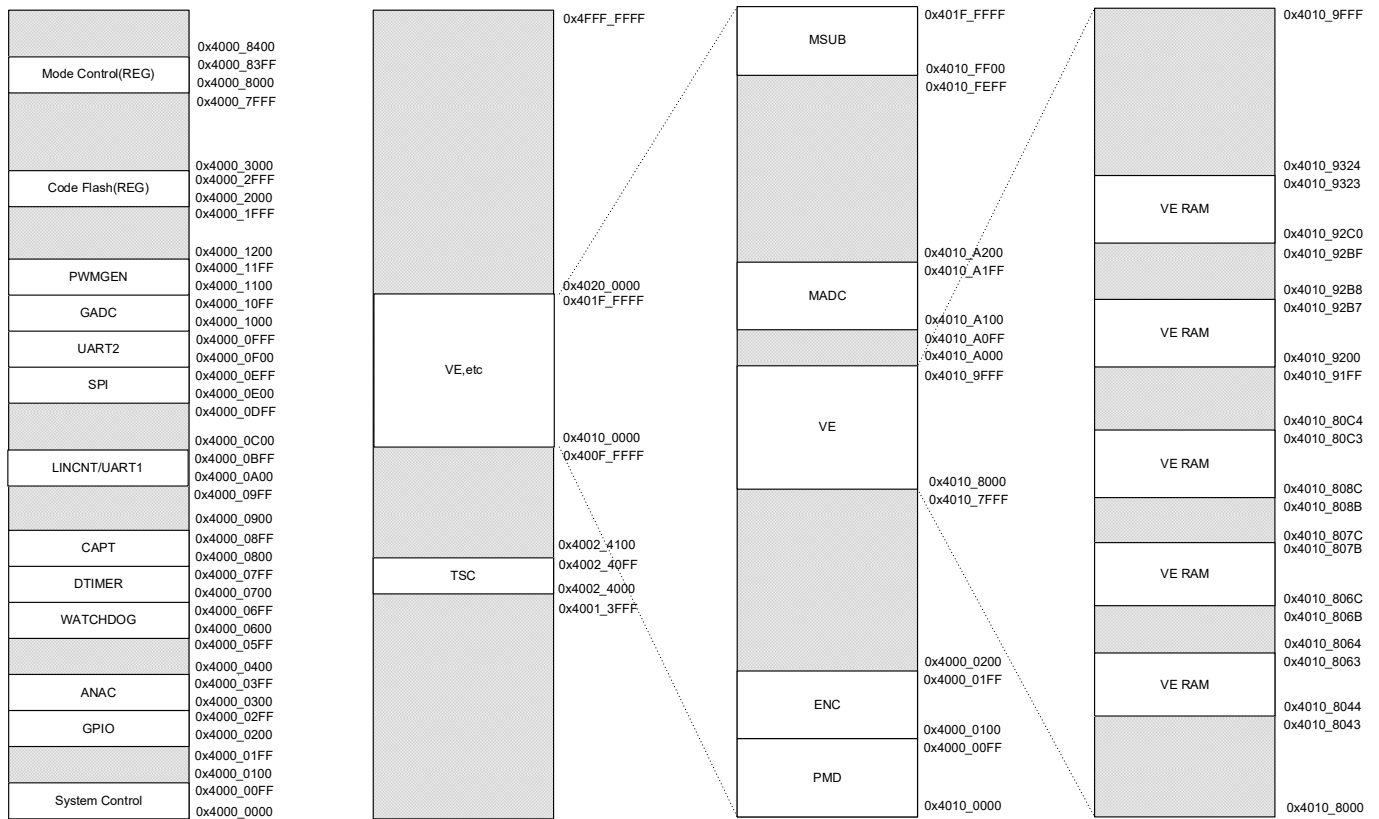
Table 9.2 PMD Function List

Category	Function	Description
PWM output	Resolution	The count resolution for PWM carriers is $1/VECLK$ . The PWM frequency and duty cycle are programmed with 15-bit values.
	PWM carrier generation	Capable of generating PWM carriers with a frequency between 0.06 kHz and 117.18 kHz (when $VECLK=60$ MHz) and with a 15-bit amplitude <ul style="list-style-type: none"> <li>Four types of carrier waveform (triangular, sawtooth, inverted triangular, and inverted sawtooth)</li> <li>Selectable carrier waveforms for each phase</li> <li>Capable of producing phase differences between the base carrier and each of the U, V, and W phases independently.</li> </ul>
	Three-phase PWM generation	Three-phase PWM waveforms are generated by comparing the PWM carriers and the programmed duty cycles. The duty cycles of the three phases can be the same or independently programmed.
	Commutation control	Each of the six output ports can be programmed to be driven to the PWM output level, logic High, or logic Low. PWM waveforms can be generated independently for each of the three phases with a common PWM carrier (three-phase complementary PWM).
AD conversion triggering	Synchronous ADC trigger generation	Generates MADC synchronous trigger signals to start AD conversion at any timing synchronous to the PWM carrier
Protection functions	Protection control	Turns off the PWM outputs in response to a protection input signal <ul style="list-style-type: none"> <li>Provides two types of protection control (EMG and OVV)</li> </ul>
	Dead-time control	Inserts a dead-time period to prevent a short-circuit between high-side and low-side switches (U/X, V/Y, W/Z) during their switching when generating three-phase complementary PWM signals
Register buffering	–	Double- and triple-buffered registers allow the PWM period, duty cycles, ADC trigger timing, and the commutation control settings for the six output ports to be changed dynamically. <ul style="list-style-type: none"> <li>The update timing of execute buffers is selectable from: 1) asynchronous, 2) the center of the PWM period, 3) the end of the PWM period, and 4) the center and end of the PWM period.</li> <li>The update timing of intermediate buffers is selectable from: 1) asynchronous, 2) the center of the PWM period, 3) the end of the PWM period, 4) the 1/4 point of the PWM period, 5) the 3/4 point of the PWM period, 6) center and end point of the PWM period, and 7) the 1/4 and 3/4 point of the PWM period.</li> </ul>
Interrupt requests	PWM interrupt (INTPWM)	Capable of generating interrupt requests synchronous to a PWM waveform <ul style="list-style-type: none"> <li>The interrupt timing is selectable from the center and end of the base carrier.</li> <li>The interval between PWM interrupts is selectable from one-half, one, two, and four PWM periods.</li> <li>Capable of enabling and disabling the decimation of synchronous MADC triggers and buffer updates when interrupts are decimated</li> </ul>
	EMG interrupt (INTEMC)	Interrupt request generated in the event of a protection event via an EMG input
	OVV interrupt (INTOVV)	Interrupt request generated in the event of a protection event via an OVV input
Debug output	–	Capable of monitoring the operating timing of the motor-related peripherals via an output port <ul style="list-style-type: none"> <li>Monitor timing of the synchronous MADC trigger output from the PMD</li> <li>Monitor timing of interrupt requests from motor-related peripherals</li> <li>Monitor MADC conversion</li> <li>Monitor timing of VE task transitions</li> <li>Monitor ENC internal signal.</li> </ul>

**9.9. Memory Map**



**Figure 9.9.1 Memory Map (Normal/Debug Mode)**



**Figure 9.9.2 Memory Map (SPECIAL FUNCTION REGISTERS)**

## 10. Electrical Characteristics

### 10.1. Absolute Maximum Ratings

Table 10.1.1 Absolute Maximum Ratings

Characteristic	Symbol	Pin(s)	Rating	Unit
Supply voltage	Vbat	VBAT	-0.3 to +40	V
	Vcp	VCP	-0.3 to +40	
	Vcc	VCC, VREF	-0.3 to +6	
	Vdd	VDD	-0.3 to +2.1	
Ground-to-ground voltage differential	Vgnd	GNDA, GNDD, GNDDP, GNDLIN	-0.3 to +0.3	
Input voltage	Vin1	LIN	-27 to +40 (Note1)	V
	Vin2	VBATD,	-0.3 to Vcp+0.3 (40 V max)	
	Vin3	PWMIN	-4 to Vbat+0.3 (40 V max) (Note2), (Note5)	
	Vin4	VMON	-0.3 to Vbat+0.3 (40 V max)	
	Vin5	U, V, W, COM, SLC	-2.5 to Vcp+0.3 (40 V max) (Note3), (Note6)	
	Vin6	RSH, RSL	-2 to Vbat+0.3 (40 V max) (Note5)	
	Vin7	TEST, MD0, MD1, RESETN GPIO_x, XIN	-0.3 to Vcc+0.3 (6 V max)	
Output voltage	Vout1	LIN	-27 to +40 (Note1)	V
	Vout2	GHU, GHV, GHW GLU, GLV, GLW	-0.3 to Vcp+0.3 (40 V max) (Note4)	
	Vout3	CP1H, CP2H	-0.3 to Vcp+0.3 (40 V max)	
	Vout4	CP1L, CP2L	-0.3 to Vbat+0.3 (40 V max)	
	Vout5	XOUT, RESETN GPIO_x	-0.3 to Vcc+0.3 (6 V max)	
Operating temperature	Ta	-	-40 to +150	°C
Storage temperature	Tstg	-	-55 to +150	°C

Note:

- None of the absolute maximum ratings must be exceeded even instantaneously. Exposure to stress exceeding absolute maximum ratings might cause permanent destruction or degradation of an IC and adversely affect other components.

Ensure that none of the absolute maximum ratings is exceeded under any operating conditions.

Use the IC within the specified operating ranges.

- At outside the  $\pm 18$ -V range, there is a limit to the period during which the TB9M030FG may be exposed to such conditions:  $\leq 90$  minutes at 18 to 28 V and  $\leq 400$  ms at 28 to 40 V

Note1: VBAT = 6.0 to 18 V

Note2: Vin3 is the voltage applied by connecting 10k $\Omega$  minimum in series.

Note3: Vin5 is the voltage applied by connecting 22 $\Omega$  (typ. tolerance of  $\pm 5\%$ ) minimum in series to the U, V, and W terminals and 47 $\Omega$  (typ. tolerance of  $\pm 5\%$ ) minimum in series to the COM and SLC terminals.

Note4: Vout2 is the voltage applied by connecting 22 $\Omega$  (typ. tolerance of  $\pm 5\%$ ) minimum in series to the GHU, GHV, and GHW terminals and 47 $\Omega$  (typ. tolerance of  $\pm 5\%$ ) minimum in series to the GLU,

GLV, and GLW terminals.

Note5: The voltage between VBAT and RSH, RSL should not exceed -0.3V to +40V.

Note6: The voltage between VCP and U, V, W, COM, SLC should not exceed -0.3V to +40V.

## 10.2. Operating Ranges

**Table 10.2.1 Operating Ranges**

Characteristic	Symbol	Rating	Unit	Remarks
Supply voltage	Vbat	18 to 27	V	Electrical characteristics are not guaranteed. (Note1)
		6 to 18		Vbat range in which electrical characteristics are guaranteed.
Operating temperature	Topr	-40 to 150	°C	Ambient temperature, Ta
		-40 to 175		Junction temperature, Tj (Note1)

Note1: Not tested in a pre-shipment test

## 10.3. Overall Electrical Characteristics

**Table 10.3 Electrical Characteristics**

VBAT=6 to 18 V, VCC=4.8 to 5.2 V, VDD=1.45 to 1.55 V, and Tj=-40 to 175°C unless otherwise noted

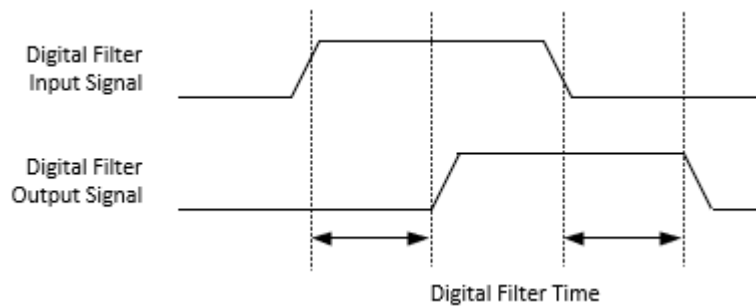
Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Standby current 1	Istby1	VBAT VBATD VMON	Standby mode LFCLK OFF VBAT=VBATD=12 V, Ta=25°C	-	-	20	μA
Standby current 2	Istby2		Standby mode LFCLK OFF VBAT=VBATD=12 V, Ta=70°C Guaranteed by design	-	-	90	μA

**10.4. Reset Generator and Standby Time**

**Table 10.4 Electrical Characteristics**

Vbat=6 to 18 V, VCC=4.8 to 5.2 V, VDD=1.45 to 1.55 V, and Tj=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Low-level reset output voltage	Voutlrst	RESETN	Iout=+5 mA	0	-	0.4	V
High-level input threshold voltage	Vihrst		-	VCC ×0.75	-	-	V
Low-level input threshold voltage	Vilrst		-	-	-	VCC ×0.25	V
Input noise filter (analog)	Tfilrst		(See Figure 10.4.1)	10	20	40	µs
Input pull-up resistor	Rpullup		Between the VCC and RESETN pins	30	50	100	kΩ
WATCHDOG reset time	Twdt	-	Time from a WATCHDOG reset request to an internal reset release (except when the flash memory is busy)	-	70	-	µs
Power-on reset settling time	Trst1	-	Time from recovery from a VCC undervoltage condition to a CPU reset release	-	-	8	ms
Recovery settling time	Trst2	-	Time from when a wake-up request is detected to when a CPU reset is released after the LDO15V stabilizes	-	-	2	ms
External oscillator settling time	Trst3	-	Time from when the oscillator is started by software to when it settles to a steady state (when a 16-MHz CERALOCK ceramic resonator from Murata is used)	-	-	1	ms
PLL settling time	Tpll	-	-	-	-	140	µs



**Figure 10.4.1 Measurement of Digital Filter Time**

## 10.5. 5-V Regulator (LDO5V)

Table 10.5 Electrical Characteristics

VBAT=6 to 18 V, VCC=4.8 to 5.2 V, VDD=1.45 to 1.55 V, and Tj=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
VCC output voltage 1	Vcc1	VCC	Iload = -10 $\mu$ A to -135 mA (Total VCC/VDD current consumption of maximum self consumption current inside the IC and external load current (max. 60mA)) Tj = -40 to 150°C	4.9	5.0	5.1	V
VCC output voltage 3	Vcc3		Iload = -10 $\mu$ A to -135 mA (Total VCC/VDD current consumption of maximum self consumption current inside the IC and external load current (max. 60mA)) Tj = 150 to 175°C	4.8	5.0	5.2	V
Current limit 1	Ilimit1		VCC $\geq$ 4.0 V	-850	-475	-250	mA
Current limit 2	Ilimit2		VCC $\leq$ 3.0 V	-250	-112	-10	mA
Dropout voltage	Vdrop		Vbat = 4.8 V, Iload = -120 mA < (-5mA + current consumed by LDO5V)	-	0.2	0.45	V
Undervoltage detection voltage 1	Vrst1		VCC falling (UV_VCC)	4.0	-	4.35	V
Undervoltage release voltage 1	Vrstr1		VCC rising (UV_VCC)	4.2	-	4.75	V
Undervoltage detection voltage 3	Vrst3		VCC falling (POR5V)	3.07	3.45	3.83	V
Undervoltage release voltage 3	Vrstr3	VCC rising (POR5V)	3.22	3.60	3.98	V	

Note:

- Connect a capacitor of 1.0  $\mu$ F or more as close as possible to the VCC pin.
- The current limit at VCC in Standby mode is Current Ilimit2.

## 10.6. 1.5-Voltage Regulator (LDO15V)

Table 10.6 Electrical Characteristics

VBAT=6 to 18 V, VCC=4.8 to 5.2 V, VDD=1.45 to 1.55 V, and Tj=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
VDD output voltage	Vdd	VDD	Iload = -10 $\mu$ A to -60 mA + (Total VDD current consumption of maximum self consumption current inside the IC and external load current (max. 1mA))	1.45	1.5	1.55	V
Current limit 3	Ilimit3		-	-250	-150	-70	mA
Undervoltage detection voltage 2	Vrst2		VDD falling	1.3	-	1.4	V
Undervoltage release voltage 2	Vrstr2		VDD rising	1.35	-	1.45	V
Overvoltage detection/release voltage	Vddov		-	1.55	-	1.65	V

Note: Connect a capacitor of 2.2  $\mu$ F or more as close as possible to the VDD pin.

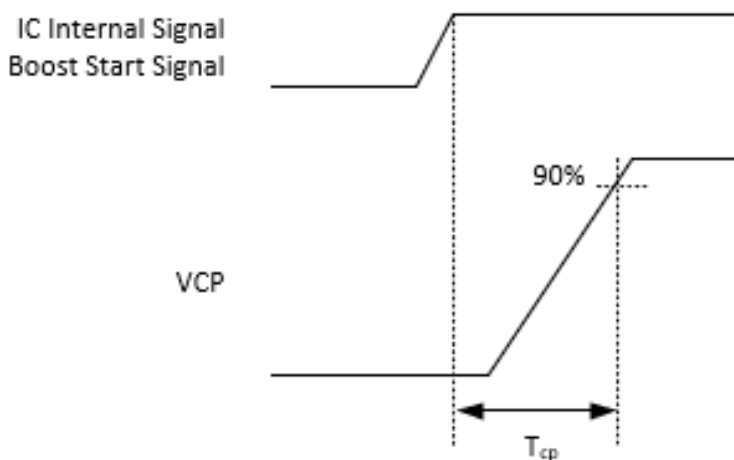
**10.7. Charge Pump**

**Table 10.7 Electrical Characteristics**

VBAT=6 to 18 V, VCC=4.8 to 5.2 V, VDD=1.45 to 1.55 V, and Tj=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Output voltage 1	Vcp1	VCP	VBAT = 6 V, Iload = -10 μA to -31.8mA, Cload=15,000 pF	VBAT +6.5	-	-	V
Output voltage 2	Vcp2		VBAT = 8 to 18V, Iload = -10 μA to -31.8mA, Cload=15,000 pF	VBAT +10	VBAT +12	VBAT +14	V
Output voltage 3	Vcp3		VBAT=5 V, Iload = -10 μA to -13.8mA, Cload=5,500 pF	VBAT +6.5	VBAT +8.7	-	V
Boost overvoltage detection threshold voltage	Vcplim1	VCP	-	31	33	35	V
Boost overvoltage recovery threshold voltage	Vcplim_r1		-	29.5	31.5	33.5	V
Boost disable threshold voltage 1	Vcpstop1	VBAT	-	27	28.5	30	V
Boost enable threshold voltage 1	Vcpstop_r1		-	26	27.5	29	V
Boost disable threshold voltage 2	Vcpstop2	VCP	-	34	36	38	V
Boost enable threshold voltage 2	Vcpstop_r2		-	32	34	36	V
Boost frequency	Fcp	-	-	237.5	250	262.5	kHz
Rise time	Tcp	VCP	Time from when the boost start signal is asserted to when the output voltage reaches 90% of Vcp	-	-	1	ms

Note: Enable the charge pump when VBAT ≥ 5 V.



**Figure 10.7.1 Measurement of Rise Time**

## 10.8. Oscillator

Table 10.8 Electrical Characteristics

V<sub>BAT</sub>=6 to 18 V, V<sub>CC</sub>=4.8 to 5.2 V, V<sub>DD</sub>=1.45 to 1.55 V, and T<sub>j</sub>=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Oscillation frequency 1 (HFCLK)	Fosc1	-	-	19	20	21	MHz
Oscillation frequency 2 (LFCLK)	Fosc2	-	-	24	32	40	kHz
Oscillation frequency 3 (XCLK)	Fosc3	XIN XOUT	Supported external ceramic resonators or crystal oscillator parameters	16 (Note1)	-	20 (Note1)	MHz

Note1: Typical value of an external oscillator.

Contact the manufacturer of external components for the matching with XCLK.

Toshiba has tested the CSTNE16M0VH3C000R0 and CSTNE20M0VH3C000R0 and confirmed that they operate with the XCLK properly.

## 10.9. 12-bit ADC

Table 10.9 Electrical Characteristics

V<sub>BAT</sub>=6 to 18 V, V<sub>CC</sub>=4.8 to 5.2 V, V<sub>DD</sub>=1.45 to 1.55 V, and T<sub>j</sub>=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Conversion time 1	Tconv1	-	Settling time + Conversion time MADCCLK=30MHz	-	1.33	-	μs
INL1	Inl1	-	Guaranteed by design	-2	-	2	LSB
DNL1	Dnl1	-	Guaranteed by design	-1	-	2	LSB
Total error 1	Err_total1	-	Guaranteed by design	-6	-	6	LSB
Input voltage division ratio	Ratio_r1	VBATD	Buffer input voltage division ratio Input voltage range: 6 to 27 V	0.095	0.1	0.105	Times

Note: When the 12bit ADC is enabled, the pull-down resistors for reducing the input voltage have a value of 250 kΩ typical.

## 10.10. 10-bit ADC

Table 10.10 Electrical Characteristics

V<sub>BAT</sub>=6 to 18 V, V<sub>CC</sub>=4.8 to 5.2 V, V<sub>DD</sub>=1.45 to 1.55 V, and T<sub>j</sub>=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Conversion time 2	Tconv2	-	Settling time + Conversion time GADCCLK=10MHz	-	16.4	-	μs
INL2	Inl2	-	-	-2.5	-	2.5	LSB
DNL2	Dnl2	-	-	-1.5	-	1.5	LSB
Total error 2	Err_total2	-	-	-3	-	3	LSB
Input voltage division ratio	Ratio_r2	V <sub>MON</sub>	Buffer input voltage division ratio Input voltage range: 6 to 27 V	0.095	0.1	0.105	Times
Amplifier error	Err_amp2	-	Buffer amplifier input-output differential Input voltage range between 0.2 V and supply voltage	-10	-	10	mV

## 10.11. Gate Driver Circuit (PREDRV)

### Table 10.11 Electrical Characteristics

V<sub>BAT</sub>=6 to 18 V, V<sub>CC</sub>=4.8 to 5.2 V, V<sub>DD</sub>=1.45 to 1.55 V, and T<sub>j</sub>=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Output voltage HH1	Vohh1	GHU GHV GHW	High-side high-level output Referenced to the source of FETs V <sub>BAT</sub> ≥ 8 V, I <sub>cp</sub> =31.8 mA C <sub>load</sub> =15,000 pF (Note1)	8	10	14	V
Output voltage HH2	Vohh2		High-side high-level output Referenced to the source of FETs V <sub>BAT</sub> ≥ 6V, I <sub>cp</sub> =31.8 mA C <sub>load</sub> =15,000 pF (Note1)	4.5	-	14	V
Output voltage HH3	Vohh3		High-side high-level output Referenced to the source of FETs V <sub>BAT</sub> ≥ 4.8V, I <sub>cp</sub> =31.8 mA C <sub>load</sub> =15,000 pF (Note1) Process variations are not considered. (Only temperature variations are considered.) Guaranteed by design	3.8	-	14	V
Output voltage HH4	Vohh4		High-side high-level output Referenced to the source of FETs V <sub>BAT</sub> ≥ 5V, I <sub>cp</sub> =13.8mA C <sub>load</sub> =5,500 pF (Note2)	4.5	-	14	V
Output voltage LH	Volh		High-side low-level output I <sub>gx</sub> =100 μA * The VB values of 4.8 to 6 V and 18 to 27 V are guaranteed by design.	-0.1	-	0.1	V
Output voltage HL1	Vohl1	GLU GLV GLW	Low-side high-level output Referenced to the source of FETs V <sub>BAT</sub> ≥ 8 V, I <sub>cp</sub> =31.8 mA C <sub>load</sub> =15,000 pF (Note1)	8	10	14	V
Output voltage HL2	Vohl2		Low-side high-level output Referenced to the source of FETs V <sub>BAT</sub> ≥ 6V, I <sub>cp</sub> =31.8 mA C <sub>load</sub> =15,000 pF (Note1)	4.5	-	14	V
Output voltage HL3	Vohl3		Low-side high-level output Referenced to the source of FETs V <sub>BAT</sub> ≥ 4.8 V, I <sub>cp</sub> =31.8 mA C <sub>load</sub> =15,000 pF (Note1) Process variations are not considered. (Only temperature variations are considered.) Guaranteed by design	3.8	-	14	V
Output voltage HL4	Vohl4		Low-side high-level output Referenced to the source of FETs Guaranteed by design: V <sub>BAT</sub> ≥ 5 V	4.5	-	14	V
Output voltage LL	Volll		Low-side low-level output I <sub>gx</sub> =100 μA * The V <sub>BAT</sub> values of 4.8 to 6 V and 18 to 27 V are guaranteed by design.	-0.1	-	0.1	V

Note1: Although the PREDRV specifications are prescribed at an I<sub>cp</sub> of 31.8 mA (i.e., a load current of the charge pump assuming that six FETs with a total C<sub>load</sub> of 15,000 pF are driven at a PWM frequency of 20 kHz), these output voltage characteristics are tested with a load current (I<sub>gx</sub>) of -100 μA.

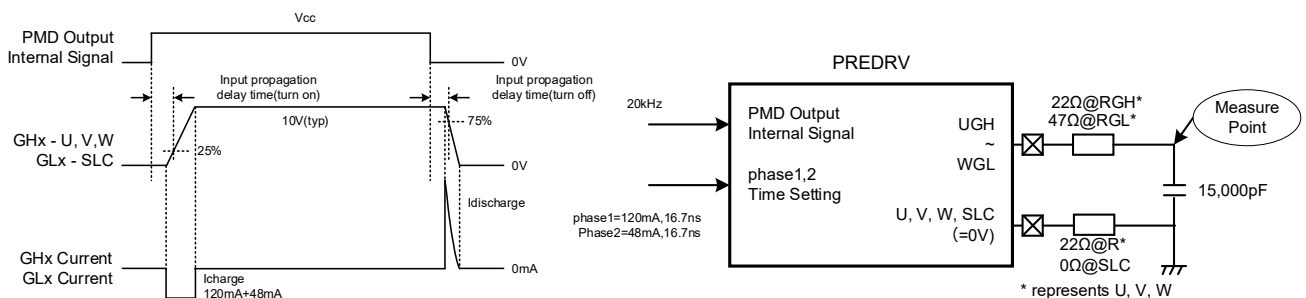
Note2: Although the PREDRV specifications are prescribed at an I<sub>cp</sub> of 13.8 mA (i.e., a load current of the charge pump assuming that six FETs with a total C<sub>load</sub> of 5,500 pF are driven at a PWM frequency of 20 kHz), this output voltage characteristic is tested with a load current (I<sub>gx</sub>) of -100 μA.

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Phase-1 charge current 1	I <sub>phase1c1</sub>	GHU GHV GHW GLU GLV GLW	Minimum setting for Phase 1: [PREDRVSRCR]<IPHASE1> = 0000 (min.0mA) Common to the high-side and low-side gate drivers	-0.1	0	0.1	mA
Phase-1 charge current 2	I <sub>phase1c2</sub>		[PREDRVSRCR]<IPHASE1> = 1000 (15mA) Common to the high-side and low-side gate drivers	-20.25	-15	-9.75	mA
Phase-1 charge current 3	I <sub>phase1c3</sub>		[PREDRVSRCR]<IPHASE1> = 1001 (30mA) Common to the high-side and low-side gate drivers	-40.5	-30	-19.5	mA
Phase-1 charge current 4	I <sub>phase1c4</sub>		[PREDRVSRCR]<IPHASE1> = 1011 (60mA) Common to the high-side and low-side gate drivers	-81	-60	-39	mA
Phase-1 charge current 5	I <sub>phase1c5</sub>		Maximum setting Phase 1: [PREDRVSRCR]<IPHASE1> = 1111 (max.120mA) Common to the high-side and low-side gate drivers	-162	-120	-78	mA

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Phase-2 high-side current 1	I <sub>phase2h1</sub>	GHU GHV GHW GLU GLV GLW	Minimum high-side setting for Phase 2: [PREDRVSRCR]<IPHASE2H> = 00000 (Min.1mA)	-1.45	-1	-0.55	mA
Phase-2 high-side current 2	I <sub>phase2h2</sub>		[PREDRVSRCR]<IPHASE2H> = 00001 (2mA)	-2.7	-2	-1.3	mA
Phase-2 high-side current 3	I <sub>phase2h3</sub>		[PREDRVSRCR]<IPHASE2H> = 00011 (4mA)	-5.4	-4	-2.6	mA
Phase-2 high-side current 4	I <sub>phase2h4</sub>		[PREDRVSRCR]<IPHASE2H> = 00111 (8mA)	-10.8	-8	-5.2	mA
Phase-2 high-side current 5	I <sub>phase2h5</sub>		[PREDRVSRCR]<IPHASE2H> = 01111 (16mA)	-21.6	-16	-10.4	mA
Phase-2 high-side current 6	I <sub>phase2h6</sub>		[PREDRVSRCR]<IPHASE2H> = 10000 (18mA)	-24.3	-18	-11.7	mA
Phase-2 high-side current 7	I <sub>phase2h7</sub>		[PREDRVSRCR]<IPHASE2H> = 10001 (20mA)	-27	-20	-13	mA
Phase-2 high-side current 8	I <sub>phase2h8</sub>		[PREDRVSRCR]<IPHASE2H> = 10011 (24mA)	-32.4	-24	-15.6	mA
Phase-2 high-side current 9	I <sub>phase2h9</sub>		[PREDRVSRCR]<IPHASE2H> = 10111 (32mA)	-43.2	-32	-20.8	mA
Phase-2 high-side current 10	I <sub>phase2h10</sub>		Maximum high-side setting for Phase 2: [PREDRVSRCR]<IPHASE2H> = 11111 (Max.48mA)	-64.8	-48	-31.2	mA

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Phase-2 low-side current 1	Iphase211	GHU GHV GHW GLU GLV GLW	Minimum low-side setting for Phase 2: [PREDRVSRCCR]<IPHASE2L> = 00000 (Min.1mA)	-1.45	-1	-0.55	mA
Phase-2 low-side current 2	Iphase212		[PREDRVSRCCR]<IPHASE2L> = 00001 (2mA)	-2.7	-2	-1.3	mA
Phase-2 low-side current 3	Iphase213		[PREDRVSRCCR]<IPHASE2L> = 00011 (4mA)	-5.4	-4	-2.6	mA
Phase-2 low-side current 4	Iphase214		[PREDRVSRCCR]<IPHASE2L> = 00111 (8mA)	-10.8	-8	-5.2	mA
Phase-2 low-side current 5	Iphase215		[PREDRVSRCCR]<IPHASE2L> = 01111 (16mA)	-21.6	-16	-10.4	mA
Phase-2 low-side current 6	Iphase216		[PREDRVSRCCR]<IPHASE2L> = 10000 (18mA)	-24.3	-18	-11.7	mA
Phase-2 low-side current 7	Iphase217		[PREDRVSRCCR]<IPHASE2L> = 10001 (20mA)	-27	-20	-13	mA
Phase-2 low-side current 8	Iphase218		[PREDRVSRCCR]<IPHASE2L> = 10011 (24mA)	-32.4	-24	-15.6	mA
Phase-2 low-side current 9	Iphase219		[PREDRVSRCCR]<IPHASE2L> = 10111 (32mA)	-43.2	-32	-20.8	mA
Phase-2 low-side current 10	Iphase2110		Maximum low-side setting for Phase 2: [PREDRVSRCCR]<IPHASE2L> = 11111 (Max.48mA)	-64.8	-48	-31.2	mA
Phase-3 current	Iphase3		Maximum high-side setting for Phase 2: [PREDRVSRCCR]<IPHASE2H> = 11111 (Max.48mA) Maximum low-side setting for Phase 2: [PREDRVSRCCR]<IPHASE2H> = 11111 (Max.48mA) Separate to the high-side and low-side gate drivers	-226.8	-168	-109.2	mA

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Output resistor 1	Routh	GHU GHV GHW	High-side discharge side Iol=-20 mA	2	5.4	12	Ω
Output resistor 2	Routl	GLU GLV GLW	Low-side discharge side Iol=-20 mA	2	5.4	12	Ω
Input propagation delay time 1	Thondelay	GHU GHV GHW GLU GLV GLW	High-side turn-on Cload=15,000 pF, 25% of VGHx Phase-1 charge current setting register [PREDRVSRCR]<IPHASE1> = 111 (120 mA max) Phase -2 charge current setting register [PREDRVSRCR]<IPHASE2H> = 11111 (48 mA max) Total: 168 mA U, V, W, SLC= 0 V (See Figure 10.11.1.)	-	0.4	1.0	μs
Input propagation delay time 2	Thoffdelay		High-side turn-off Cload=15,000 pF, 75% of VGHx U, V, W are 0V through RU,RV,RW (External parts: RGHU,RGHV,RGHW = RU,RV,RW = 22Ω, C=15,000 pF) Observed outside of RGHU,RGHV,RGHW (See Figure 10.11.1.)	-	0.3	1.0	μs
Input propagation delay time 3	Tlondelay		Low-side turn-on Cload=15,000 pF, 25% of VGLx Phase-1 charge current setting register [PREDRVSRCR]<IPHASE1>= 111 (120 mA max) Phase -2 charge current setting register [PREDRVSRCR]<IPHASE2L>= 11111 (48 mA max) Total: 168 mA 10%-90% U, V, W, SLC= 0 V (See Figure 10.11.1.)	-	0.4	1.0	μs
Input propagation delay time 4	Tloffdelay		Low-side turn-off Cload=15,000 pF, 75% of VGLx U, V, W, SLC= 0 V (External parts: R=47Ω, C=15,000 pF) Observed outside the 47Ω resistor (See Figure 10.11.1.)	-	0.3	1.0	μs



**Figure 10.11.1 Measurement of Input Propagation Delay Times and Measurement circuit diagram**

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Initial Phase-1 period 1	Tphase1i1	-	VECLK=60MHz [MCSSDRVPTSR]<TPHASE1>: 000000 (0 ns, minimum setting) Guaranteed by design	-	0	-	ns
Initial Phase-1 period 2	Tphase1i2	-	VECLK=60MHz [MCSSDRVPTSR]<TPHASE1>: 000001 (16.7 ns)	15.8	16.7	17.5	ns
Initial Phase-1 period 3	Tphase1i3	-	VECLK=60MHz [MCSSDRVPTSR]<TPHASE1>: 000011 (50 ns)	47.5	50	52.5	ns
Initial Phase-1 period 4	Tphase1i4	-	VECLK=60MHz [MCSSDRVPTSR]<TPHASE1>: 000111 (116.7 ns)	110.9	116.7	122.5	ns
Initial Phase-1 period 5	Tphase1i5	-	VECLK=60MHz [MCSSDRVPTSR]<TPHASE1>: 001111 (250 ns)	237.5	250	262.5	ns
Initial Phase-1 period 6	Tphase1i6	-	VECLK=60MHz [MCSSDRVPTSR]<TPHASE1>: 011111 (516.7 ns)	490.9	516.7	542.5	ns
Initial Phase-1 period 7	Tphase1i7	-	VECLK=60MHz [MCSSDRVPTSR]<TPHASE1>: 111111 (1050 ns, maximum setting)	997.5	1050	1102.5	ns

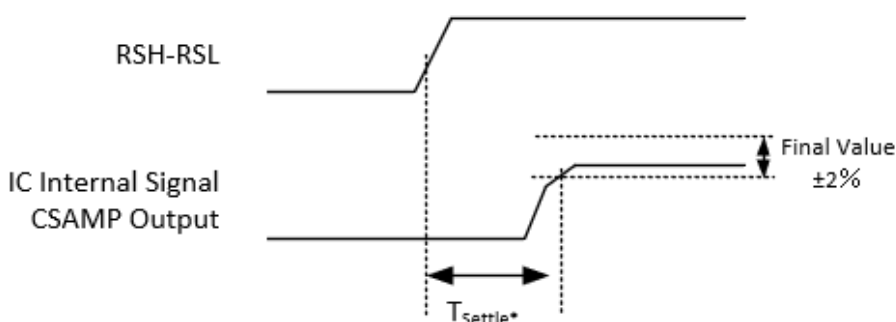
Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Phase-2 target period 1	Tphase2t1	-	VECLK=60MHz [MCSSDRVPTSR]<TPHASE2>: 0000000 (0 ns, minimum setting) Guaranteed by design	-	0	-	ns
Phase-2 target period 2	Tphase2t2	-	VECLK=60MHz [MCSSDRVPTSR]<TPHASE2>: 0000001 (16.7 ns)	15.8	16.7	17.5	ns
Phase-2 target period 3	Tphase2t3	-	VECLK=60MHz [MCSSDRVPTSR]<TPHASE2>: 0000011 (50 ns)	47.5	50	52.5	ns
Phase-2 target period 4	Tphase2t4	-	VECLK=60MHz [MCSSDRVPTSR]<TPHASE2>: 0000111 (116.7 ns)	110.9	116.7	122.5	ns
Phase-2 target period 5	Tphase2t5	-	VECLK=60MHz [MCSSDRVPTSR]<TPHASE2>: 0001111 (250 ns)	237.5	250	262.5	ns
Phase-2 target period 6	Tphase2t6	-	VECLK=60MHz [MCSSDRVPTSR]<TPHASE2>: 0011111 (516.7 ns)	490.9	516.7	542.5	ns
Phase-2 target period 7	Tphase2t7	-	VECLK=60MHz [MCSSDRVPTSR]<TPHASE2>: 0111111 (1050 ns)	997.5	1050	1102.5	ns
Phase-2 target period 8	Tphase2t8	-	VECLK=60MHz [MCSSDRVPTSR]<TPHASE2>: 1111111 (2116.7 ns)	2010.9	2116.7	2222.5	ns
Dead time	Tdead	-	Dead-Time register: [DTR]<DTR[9:0]> = 0x02D When dead time is programmed to be 3 $\mu$ s. VECLK=60 MHz	2.85	3	3.15	$\mu$ s

**10.12. Current-Sense Amplifier (CSAMP)**

**Table 10.12 Electrical Characteristics**

V<sub>BAT</sub>=6 to 18 V, V<sub>CC</sub>=4.8 to 5.2 V, V<sub>DD</sub>=1.45 to 1.55 V, and T<sub>j</sub>=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Input common-mode voltage range	V <sub>comin</sub>		-	-0.2	-	2.0	V
Gain 1	Gain1	RSH RSL	Programmed to be ×5	4.95	5	5.05	Times
Gain 2	Gain2		Programmed to be ×10	9.9	10	10.1	Times
Gain 3	Gain3		Programmed to be ×15	14.85	15	15.15	Times
Gain 4	Gain4		Programmed to be ×20	19.8	20	20.2	Times
Gain 5	Gain5		Programmed to be ×40	39.6	40	40.4	Times
Gain 6	Gain6		Programmed to be ×60	59.25	59.85	60.45	Times
Output offset voltage 1	Vo <sub>offset1</sub>	-	Programmed to be V <sub>REF</sub> / 2 Gain60 after calibration, including input offset	V <sub>REF</sub> /2 -0.0606	2.5	V <sub>REF</sub> /2 +0.0606	V
Output offset voltage 2	Vo <sub>offset2</sub>	-	Programmed to be V <sub>REF</sub> / 5 Gain60 after calibration, including input offset	V <sub>REF</sub> /5 -0.0606	1.0	V <sub>REF</sub> /5 +0.0606	V
Output offset voltage 3	Vo <sub>offset3</sub>	-	Programmed to be V <sub>REF</sub> / 8 Gain60 after calibration, including input offset	V <sub>REF</sub> /8 -0.0606	0.625	V <sub>REF</sub> /8 +0.0606	V
Output offset voltage 4	Vo <sub>offset4</sub>	-	Programmed to be V <sub>REF</sub> / 10 Gain60 after calibration, including input offset	V <sub>REF</sub> /10 -0.0606	0.5	V <sub>REF</sub> /10 +0.0606	V
Settling time 1	T <sub>settle1</sub>	-	Gain = 5, output voltage: 2.5V ↔ 3.5 V No external input filter Time required for the output voltage to settle to within ±2% of its final value Guaranteed by design (See Figure 10.12.1)	-	-	0.5	μs
Settling time 2	T <sub>settle2</sub>	-	Gain = 20, output voltage: 2.5V ↔ 3.5 V No external input filter Time required for the output voltage to settle to within ±2% of its final value Guaranteed by design (See Figure 10.12.1)	-	-	0.8	μs
Settling time 3	T <sub>settle3</sub>	-	Gain = 60, output voltage: 2.5V ↔ 3.5 V No external input filter Time required for the output voltage to settle to within ±2% of its final value Guaranteed by design (See Figure 10.12.1)	-	-	1.7	μs
High-level output voltage	V <sub>oh_amp</sub>	-	-	V <sub>CC</sub> -0.3	-	-	V
Low-level output voltage	V <sub>ol_amp</sub>	-	-	-	-	0.3	V



**Figure 10.12.1 Measurement of Settling Time**

**10.13. Current Clamp Comparator (CLCMP)**

**Table 10.13 Electrical Characteristics**

VBAT=6 to 18 V, VCC=4.8 to 5.2 V, VDD=1.45 to 1.55 V, and Tj=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Current-limiting voltage threshold 1	Vlimit1	-	[CLCMPSR]<CLCMPTV> = 00000 Minimum setting	-39	0	39	mV
Current-limiting voltage threshold 2	Vlimit2	-	[CLCMPSR]<CLCMPTV> = 00001	$VCC \times \frac{1}{32} - 0.039$	0.156	$VCC \times \frac{1}{32} + 0.039$	V
Current-limiting voltage threshold 3	Vlimit3	-	[CLCMPSR]<CLCMPTV> = 00011	$VCC \times \frac{3}{32} - 0.039$	0.468	$VCC \times \frac{3}{32} + 0.039$	V
Current-limiting voltage threshold 4	Vlimit4	-	[CLCMPSR]<CLCMPTV> = 00111	$VCC \times \frac{7}{32} - 0.039$	1.094	$VCC \times \frac{7}{32} + 0.039$	V
Current-limiting voltage threshold 5	Vlimit5	-	[CLCMPSR]<CLCMPTV> = 01111	$VCC \times \frac{15}{32} - 0.039$	2.344	$VCC \times \frac{15}{32} + 0.039$	V
Current-limiting voltage threshold 6	Vlimit6	-	[CLCMPSR]<CLCMPTV>=11111 Maximum setting	$VCC \times \frac{31}{32} - 0.039$	4.844	$VCC \times \frac{31}{32} + 0.039$	V
Digital filtering time 1	Tclfil1	-	OVV Control register: [OVVCR]<OVVCNT[4:0]> = 0x0F VECLK = 60 MHz (See Figure 10.4.1)	3.8	4.0	4.2	μs
Digital filtering time 4	Tclfil4	-	OVV Control register: [OVVCR]<OVVCNT[4:0]> = 0x06 VECLK = 60 MHz (See Figure 10.4.1)	1.52	1.6	1.68	μs
Digital filtering time 2	Tclfil2	-	[CLCMPDFSR]<CLCMPDFS>=0 32 count at 10 MHz (See Figure 10.4.1)	3.04	3.2	3.36	μs
Digital filtering time 3	Tclfil3	-	[CLCMPDFSR]<CLCMPDFS>=1 16 count at 10 MHz (See Figure 10.4.1)	1.52	1.6	1.68	μs

Note: Digital filtering time 1 and 4 is a characteristic of a digital filter located on a different path from the one with digital filtering time 2 and digital filtering time 3.

The digital filter with digital filtering time 1 and 4 can be disabled via a special function register.

## 10.14. Overcurrent Detection Comparator (OCCMP)

Table 10.14 Electrical Characteristics

Vbat=6 to 18 V, VCC=4.8 to 5.2 V, VDD=1.45 to 1.55 V, and Tj=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Voltage threshold 1 for overcurrent detection	Voc1	-	[OCCMPSR]<OCCMPTV> = 0000 Minimum setting	-39	0	39	mV
Voltage threshold 2 for overcurrent detection	Voc2	-	[OCCMPSR]<OCCMPTV> = 0001	$VCC \times \frac{1}{16}$ -0.039	0.313	$VCC \times \frac{1}{16+0.039}$	V
Voltage threshold 3 for overcurrent detection	Voc3	-	[OCCMPSR]<OCCMPTV> = 0011	$VCC \times \frac{3}{16}$ -0.039	0.9375	$VCC \times \frac{3}{16+0.039}$	V
Voltage threshold 4 for overcurrent detection	Voc4	-	[OCCMPSR]<OCCMPTV> = 0111	$VCC \times \frac{7}{16}$ -0.039	2.1875	$VCC \times \frac{7}{16+0.039}$	V
Voltage threshold 5 for overcurrent detection	Voc5	-	[OCCMPSR]<OCCMPTV> = 1111 Maximum setting	$VCC \times \frac{15}{16}$ -0.039	4.6875	$VCC \times \frac{15}{16+0.039}$	V
Digital filtering time 1	Tocfil1	-	[EMGCR]<EMGCNT[4:0]> = 0x06 VECLK = 60 MHz (See Figure 10.4.1)	1.52	1.6	1.68	μs
Digital filtering time 2	Tocfil2	-	[OCCMPDFSR]<OCCMPDFS> = 0 32 count at 10 MHz (See Figure 10.4.1)	3.04	3.2	3.36	μs
Digital filtering time 3	Tocfil3	-	[OCCMPDFSR]<OCCMPDFS> = 1 16 count at 10 MHz (See Figure 10.4.1)	1.52	1.6	1.68	μs

Note: Note that the longer one of digital filtering time 1 and digital filtering time 2/3 becomes effective. The digital filter with digital filtering time 1 can be disabled via a special function register.

**10.15. Back-EMF Comparator (BEMFCMP)**

**Table 10.15 Electrical Characteristics**

VBAT=6 to 18 V, VCC=4.8 to 5.2 V, VDD=1.45 to 1.55 V, and Tj=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Input voltage range	Vinrange	U, V, W, COM	Input via a 47Ω resistor COM has half the voltage of the phase voltage.	-2.5	-	VBAT	V
Input offset voltage 1	Voffset1	U, V, W, COM	COM = 0 V COM voltage phase difference.	-80	-	80	mV
Input offset voltage 21	Voffset21		COM=VBAT/2, VBAT=6V COM voltage phase difference.	-75	-	75	mV
Input offset voltage 22	Voffset22		COM=VBAT/2, VBAT=12 V COM voltage phase difference.	-50	-	50	mV
Input offset voltage 23	Voffset23		COM=VBAT/2, VBAT=18 V COM voltage phase difference.	-75	-	75	mV
Input offset voltage 1 phase differential	Voffset1d	U, V, W	COM= 0 V, U-V, V-W, and W-U phase differentials	-99	-	99	mV
BEMFCMP filtering time	Tbemffil	-	Input Processing Control register: [INPCR]<FILVAL[6:0]> = 0x3C VECLK = 60 MHz (See Figure 10.4.1)	0.95	1	1.05	μs

**10.16. Thermal Shutdown (TSD) Circuit**

**Table 10.16 Electrical Characteristics**

VBAT=6 to 18 V, VCC=4.8 to 5.2 V, VDD=1.45 to 1.55 V, and Tj=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
TSD detection temperature	Ttsd	-	Guaranteed by design	175	190	205	°C
TSD recovery temperature	Ttsdr	-	Guaranteed by design	145	160	175	°C

Note:

- TSD is not tested in a pre-shipment test. Only an evaluation is performed.
- The TSD circuits for the 5-volt regulator (LDO5V) and the LINPHY have the same electrical characteristics.

**10.17. Open/Short Comparator (OSCMP) for External FETs**

**Table 10.17 Electrical Characteristics**

VBAT=6 to 18 V, VCC=4.8 to 5.2 V, VDD=1.45 to 1.55 V, and Tj=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Short-circuit detection threshold 1	Vos1	VBATD U, V, W SLC	VBATD-U/V/W, U/V/W-SLC [OSCMPSR]<OSCMPT> = 000 Minimum setting	10	240	460	mV
Short-circuit detection threshold 2	Vos2		VBATD-U/V/W, U/V/W-SLC [OSCMPSR]<OSCMPT> = 001	60	280	510	mV
Short-circuit detection threshold 3	Vos3		VBATD-U/V/W, U/V/W-SLC [OSCMPSR]<OSCMPT> = 010	160	380	610	mV
Short-circuit detection threshold 4	Vos4		VBATD-U/V/W, U/V/W-SLC [OSCMPSR]<OSCMPT> = 011	270	480	720	mV
Short-circuit detection threshold 5	Vos5		VBATD-U/V/W, U/V/W-SLC [OSCMPSR]<OSCMPT> = 100	360	580	830	mV
Short-circuit detection threshold 6	Vos6		VBATD-U/V/W, U/V/W-SLC [OSCMPSR]<OSCMPT> = 101	460	680	940	mV
Short-circuit detection threshold 7	Vos7		VBATD-U/V/W, U/V/W-SLC [OSCMPSR]<OSCMPT> = 110, 111 Maximum setting	560	780	1050	mV
Input current 1 (Note1)	Iosin1	U, V, W	VBAT = 12 V, U/V/W pin voltage = 12 V [OSCMPSR]<OSCMPTV> = 000 [PREDRVER]<PREDRVEN>=0	69	-	165	μA
Input current 2 (Note1)	Iosin2		VBAT = 12 V, U/V/W pin voltage = 0 V [OSCMPSR]<OSCMPTV> = 000 [PREDRVER]<PREDRVEN>=0	-5.0	-	5.0	μA
Input current 3	Iosin3	SLC	[OSCMPSR]<OSCMPTV> = 11x [PREDRVER]<PREDRVEN>=0 SLC pin sink current	-10	-	5	μA

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
<b>Digital filtering time 1 (Note2)</b>	Tshortfil5	-	Time from the rising edge of the FET "on" signal to the Vds comparator output [OSCMDFSR]<OSCMDFSR> = 00 64 count at 10 MHz (See Figure 10.4.1)	6.08	6.4	6.72	μs
<b>Digital filtering time 2 (Note2)</b>	Tshortfil6	-	Time from the rising edge of the FET "on" signal to the Vds comparator output [OSCMDFSR]<OSCMDFSR> = 01 48 count at 10 MHz (See Figure 10.4.1)	4.56	4.8	5.04	μs
<b>Digital filtering time 3 (Note2)</b>	Tshortfil7	-	Time from the rising edge of the FET "on" signal to the Vds comparator output [OSCMDFSR]<OSCMDFSR> = 10 32 count at 10 MHz (See Figure 10.4.1)	3.04	3.2	3.36	μs
<b>Digital filter time 4 (Note2)</b>	Tshortfil8	-	Time from the rising edge of the FET "on" signal to the Vds comparator output [OSCMDFSR]<OSCMDFSR> = 11 16 count at 10 MHz (See Figure 10.4.1)	1.52	1.6	1.68	μs

Note1: The U, V, and W pins are also connected to other circuit blocks. Iosin2 and Iosin2 include the amount of input currents to these circuit blocks.

Note2: The filtering time of the digital filter for the OSCMP is the longer of the time shown above (Tshortfil5 to Tshortfil8) and digital filtering time 1 (Tocfil1) of the OCCMP shown in Section 10.14.

10.18. LIN

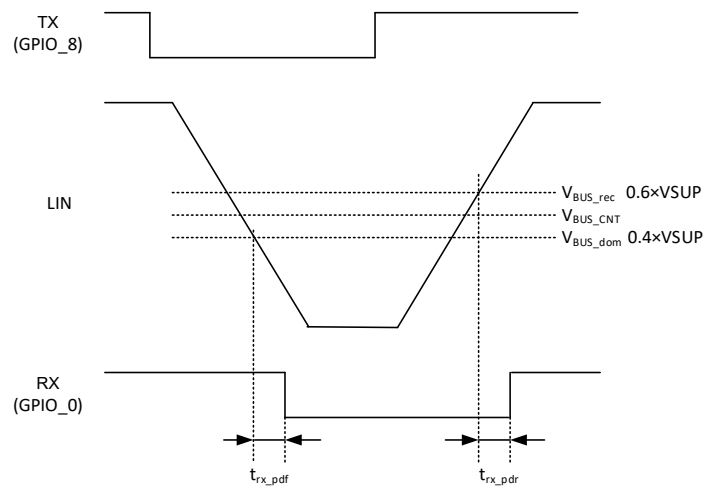
**Table 10.18 Electrical Characteristics**

VBAT=6 to 18 V, VCC=4.8 to 5.2 V, VDD=1.45 to 1.55 V, and Tj=-40 to 175°C unless otherwise noted

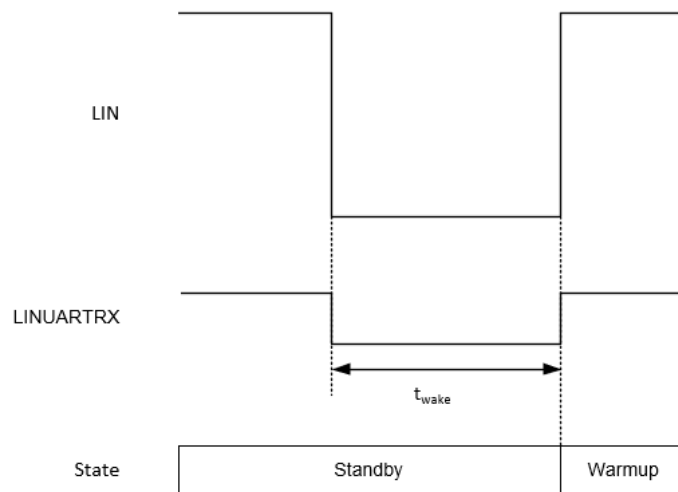
Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Supply voltage range	V <sub>VB</sub>	VBAT	ECU operating voltage range (ISO 17987-4 Param 9, SAE J2602-1)	8.0	-	18.0	V
Supply voltage range	V <sub>SUP</sub>		Note : ISO 17987-4 Param 10 and SAE J2602-1 of LIN configuration is defined 7.0V(min), IC requires 6.0V(min) on 5V regulator.	6.0	-	18.0	V
Supply voltage MAX Ratings	V <sub>SUP_NON_OP</sub>		Voltage range with in which the device is not destroyed. An optional time limit for the maximum value shall be at least 400 ms. No guarantee of correct operation. (ISO 17987-4 Param 11, SAE J2602-1)	-0.3	-	40	V
BUS MAX Ratings	V <sub>BUS_MAX_RATING</sub>	LIN	Voltage range with in which the device is not destroyed. (ISO 17987-4 Param 82) An optional time limit for the maximum value shall be at least 400 ms. No guarantee of correct operation.	-27	-	40	V
Receiver threshold voltage, recessive to dominant edge	V <sub>th_rec</sub>		Low Voltage: Recessive Input Threshold (SAE J2602-1)	0.4	-	0.53	V <sub>SUP</sub>
Receiver threshold voltage, dominant to recessive edge	V <sub>th_dom</sub>		High Voltage: Dominant Input Threshold (SAE J2602-1)	0.47	-	0.6	V <sub>SUP</sub>
BUS current limitation	I <sub>BUS_LIM</sub>		Current Limitation for Driver dominant state driver on VBUS = VBAT_max (ISO 17987-4 Param 12)	40	-	200	mA
Leakage current (dominant)	I <sub>BUS_PAS_dom</sub>		Input leakage current at the receiver incl. responder Pull Up resistance as specified in ISO 17987-4 Param 26 driver off VBUS = 0 V VBAT = 12 V (ISO 17987-4 Param 13)	-1	-	-	mA
Leakage current (recessive)	I <sub>BUS_PAS_re</sub>		Driver off 8 V < VBAT < 18 V, 8 V < VBUS < 18 V, VBUS > VBAT (ISO 17987-4 Param 14)	-	-	20	μA
Leakage current1	I <sub>BUS_NO_GND</sub>		Control unit disconnected from ground GNDDevice = VSUP 0 V < VBUS < 18 V VBAT = 12 V Loss of local ground shall not affect communication in the residual network. (ISO 17987-4 Param 15)	-1	-	1	mA
Leakage current2	I <sub>BUS_NO_BAT</sub>		VBAT disconnected VSUP = GND 0 V < VBUS < 18 V Node shall sustain the current that can flow under this condition. Bus shall remain operational under this condition. (ISO 17987-4 Param 16 and SAE J2602-1)	-	-	23	μA
Voltage of Receiver dominant state	V <sub>BUS_dom</sub>		Receiver dominant state (ISO 17987-4 Param 17) Note: ISO 17987-4 Param 17 of LINPHY configuration is not defined minimum voltage.	-27	-	0.4 x VBAT	V

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Voltage of Receiver recessive state	V <sub>BUS_rec</sub>	LIN	Receiver recessive state (ISO 17987-4 Param 18)	0.6	-	-	V <sub>SUP</sub>
Receiver center voltage	V <sub>BUS_CNT</sub>		$V_{BUS\_CNT} = (V_{th\_dom} + V_{th\_rec})/2$ (ISO 17987-4 Param 19)	0.475	-	0.525	V <sub>SUP</sub>
Receiver hysteresis	V <sub>HYS</sub>		$V_{HYS} = V_{th\_rec} - V_{th\_dom}$ (ISO 17987-4 Param 20 and SAE J2602-1)	0.07	-	0.175	V <sub>SUP</sub>
Duty cycle D1 (for worst case at 20 kbps)	D1		THRec(max) = 0,744 × V <sub>SUP</sub> ; THDom(max) = 0,581 × V <sub>SUP</sub> ; V <sub>SUP</sub> = 7,0 V to 18 V; tBIT = 50 μs; D1 = tBus_rec(min)/(2 × tBIT) (ISO 17987-4 Param 27) <PHYFBRM>=0	0.396	-	-	-
Duty cycle D2 (for worst case at 20 kbps)	D2		THRec(min) = 0,422 × V <sub>SUP</sub> ; THDom(min) = 0,284 × V <sub>SUP</sub> ; V <sub>SUP</sub> = 7,6 V to 18 V; tBIT = 50 μs; D2 = tBus_rec(max)/(2 × tBIT) (ISO 17987-4 Param 28) <PHYFBRM>=0	-	-	0.581	-
Duty cycle D3 (for worst case at 10 kbps)	D3		THRec(max) = 0,778 × V <sub>SUP</sub> ; THDom(max) = 0,616 × V <sub>SUP</sub> ; V <sub>SUP</sub> = 7,0 V to 18 V; tBIT = 96 μs; D3 = tBus_rec(min)/(2 × tBIT) (ISO 17987-4 Param 29) <PHYFBRM>=0	0.417	-	-	-
Duty cycle D4 (for worst case at 10 kbps)	D4		THRec(min) = 0,389 × V <sub>SUP</sub> ; THDom(min) = 0,251 × V <sub>SUP</sub> ; V <sub>SUP</sub> = 7,6 V to 18 V; tBIT = 96 μs; D4 = tBus_rec(max)/(2 × tBIT) (ISO 17987-4 Param 30) <PHYFBRM>=0	-	-	0.59	-
Propagation delay	t <sub>rx_pd</sub>		Propagation delay of receiver (ISO 17987-4 Param 31) • bus dominant to Rx D LOW(t <sub>rx_pdl</sub> ) • bus recessive to Rx D HIGH(t <sub>rx_pdr</sub> ) (See Figure 10.18.1)	-	-	6	μs
Receiver delay symmetry	t <sub>rx_sym</sub>		Symmetry of receiver propagation delay rising edge with respect to falling edge (ISO 17987-4 Param 32) $t_{rx\_sym} = t_{rx\_pdf} - t_{rx\_pdr}$	-2	-	2	μs
Bus pull-up resistance1	R <sub>RESPONDER</sub>		internal resistance (ISO 17987-4 Param 26, SAE J2602-1)	20	30	60	kΩ
Bus pull-up resistance2	R <sub>COMMANDER</sub>	The serial diode is mandatory. Only for valid for transceiver with integrated commander Pull Up resistance. (ISO 17987-4 Param 25, SAE J2602-1) external resistance	900	-	1100	Ω	
LIN input capacity (Guaranteed by design)	C <sub>RESPONDER</sub>	Capacitance of responder node (ISO 17987-4 Param 37) 250pF -220pF = 30pF max	-	-	30	pF	
Current consumption in sleep mode (Guaranteed by design)	I <sub>sleep_LIN</sub>	V <sub>BAT</sub>	V <sub>BAT</sub> =12V, RT=25°C, Only working LIN bus Wakeup signal Only LINPHY current consumption at [PMUSCR]<LIN_MODE>=0(Sleep Operate Mode)	-	-	3	μA

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Dominant time for bus Wakeup	$t_{WAKE}$	LIN	Wakeup pulse width from LIN bus (See Figure 10.18.2)	30	-	150	$\mu$ s
Turn off time to sleep	$t_{sleep}$	-	Turn off time from Standby or Active to Sleep mode	-	-	1	ms
Wake-up threshold voltage	$V_{BUSwk}$	LIN	Threshold voltage for Wakeup signal detection	0.4	0.5	0.6	$V_{SUP}$
ESD Susceptibility HBM1 pins LIN vs. LINGND	$V_{ESDLIN}$	LIN	IEC61000-4-2 Conducted HBM	-6	-	6	kV
ESD Susceptibility HBM3 pins LIN vs. LINGND	$V_{ESDLIN3}$		AEC-Q100-002	-6	-	6	kV



**Figure 10.18.1 Measurement of LIN Propagation delay**



**Figure 10.18.2 Measurement of LIN Dominant time for bus Wakeup**

**10.19. PWM Communication Circuit 1**

**Table 10.19 Electrical Characteristics**

VBAT=6 to 18 V, VCC=4.8 to 5.2 V, VDD=1.45 to 1.55 V, and Tj=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Output voltage H1	Vouth1	LIN	Pull-up resistor: 3.6 kΩ	VBAT-1.0	-	VBAT	V
Output voltage L1	Voutl1		Pull-up resistor: 3.6 kΩ	0	-	1.2	V

Note: Since the LIN Bus and PWM Communication Circuit 1 shares the LIN bus, they have the same electrical characteristics.

**10.20. PWM Communication Circuit 2**

**Table 10.20 Electrical Characteristics**

VBAT=6 to 18 V, VCC=4.8 to 5.2 V, VDD=1.45 to 1.55 V, and Tj=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
High-level input threshold voltage	Vinh	PWMIN	-	VCC ×0.86	VCC ×0.91	VCC ×0.96	V
Low-level input threshold voltage	Vinl		-	VCC ×0.82	VCC ×0.87	VCC ×0.92	V
Input threshold voltage hysteresis	Vinhys		Vinhys=Vinh-Vinl Guaranteed by design	-	0.2	-	V
Input current 1	lin1		Vin=18 V	-	20	36	μA
Input current 2	lin2		Vin=0 V	-1	-	1	μA
Input digital filter	Tpwmfil		The assumption is that the input digital filter for the CAPT is used. When CPUCLK is 40 MHz and divided by 128 (See Figure 10.4.1)	6.08	6.4	6.72	μs
High-level output voltage	Vouth2	GPIO_9 /PWMOUT0	Load condition: [GPOPSR]<GPIOPSx>=00: -1 mA [GPOPSR]<GPIOPSx>=01: -2 mA [GPOPSR]<GPIOPSx>=10: -4 mA [GPOPSR]<GPIOPSx>=11: -6 mA	VCC ×0.8	-	-	V
Low-level output voltage	Voutl2		Load condition: [GPOPSR]<GPIOPSx>=00: 1 mA [GPOPSR]<GPIOPSx>=01: 2 mA [GPOPSR]<GPIOPSx>=10: 4 mA [GPOPSR]<GPIOPSx>=11: 6 mA	-	-	0.5	V

**10.21. UART**

**Table 10.21 Electrical Characteristics**

VBAT=6 to 18 V, VCC=4.8 to 5.2 V, VDD=1.45 to 1.55 V, and Tj=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Data rate	Fuart	GPIO_10/U ART_RX, GPIO_11/U ART_TX	-	-	-	1	Mbps

**10.22. SPI**

**Table 10.22 Electrical Characteristics**

VBAT=6 to 18 V, VCC=4.8 to 5.2 V, VDD=1.45 to 1.55 V, and Tj=-40 to 175°C unless otherwise noted

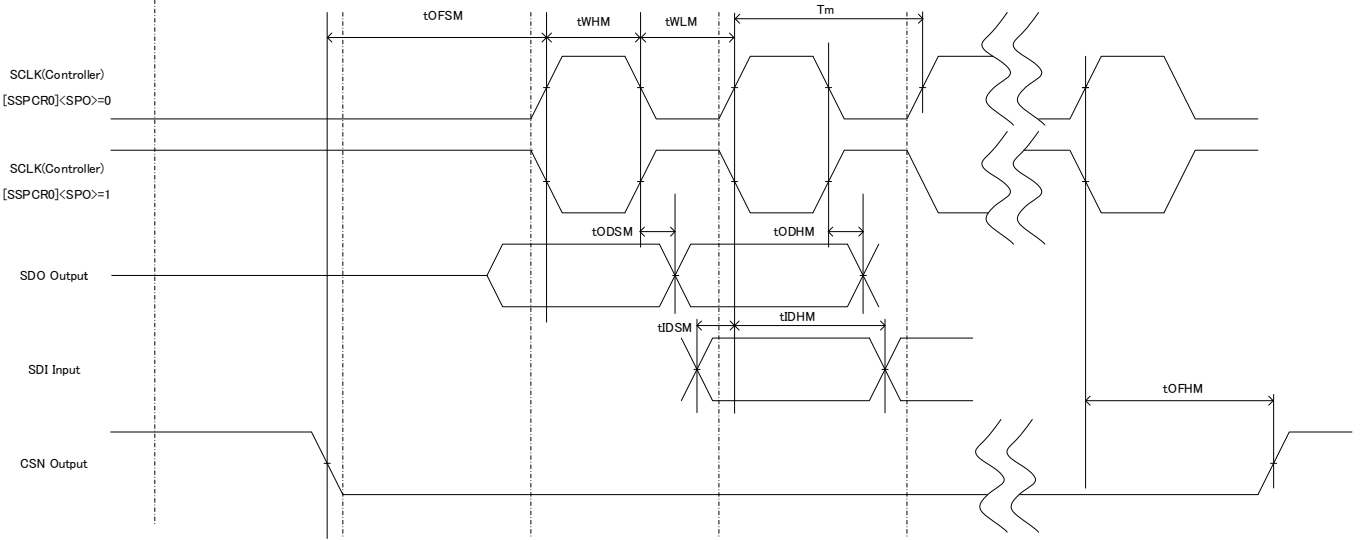
Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
SCLK period (Controller)	Tm	SCLK	(See Figure 10.22.1, Figure 10.22.2)	T×m ≥250 ns	-	-	ns
SCLK period (Target)	Ts		(See Figure 10.22.3, Figure 10.22.4)	T×n ≥1 μs	-	-	ns
SCLK Low-level pulse width in Controller mode	Twlm		(See Figure 10.22.1, Figure 10.22.2)	0.4	-	0.6	Tm
SCLK High-level pulse width in Controller mode	Twhm		(See Figure 10.22.1, Figure 10.22.2)	0.4	-	0.6	Tm
SCLK Low-level pulse width in Target mode	Twls		(See Figure 10.22.3, Figure 10.22.4)	0.4	-	0.6	Ts
SCLK High-level pulse width in Target mode	Twhs		(See Figure 10.22.3, Figure 10.22.4)	0.4	-	0.6	Ts
SCLK rise/fall to SDO valid in Controller mode	Todsm		(See Figure 10.22.1, Figure 10.22.2)	-	-	50	ns
SCLK rise/fall to SDO hold in Controller mode	Todhm		(See Figure 10.22.1, Figure 10.22.2)	-20	-	-	ns
SCLK rise/fall to SDI valid in Controller mode	Tidsm		(See Figure 10.22.1, Figure 10.22.2)	55	-	-	ns
SDI hold time from SCLK rise/fall in Controller mode	Tidhm		(See Figure 10.22.1, Figure 10.22.2)	100	-	-	ns
CSN valid to SCLK rise/fall in Controller mode	Tofsm	CSN SCLK	(See Figure 10.22.1, Figure 10.22.2)	T×m -50	-	-	ns

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
SCLK rise/fall to SDO valid in Target mode	Todss	SCLK	(See Figure 10.22.3, Figure 10.22.4)	-	-	3T+90	ns
SCLK rise/fall to SDO hold in Target mode	Todhs		(See Figure 10.22.3, Figure 10.22.4)	2T	-	-	ns
SCLK rise/fall to SDI valid in Target mode	Tidss		(See Figure 10.22.3, Figure 10.22.4)	10	-	-	ns
SCLK rise/fall edge to SDI hold in Target mode	Tidhs		(See Figure 10.22.3, Figure 10.22.4)	3T+20	-	-	ns
CSN valid to SCLK rise/fall in Target mode	Tifss	CSN SCLK	(See Figure 10.22.3, Figure 10.22.4)	T×n -20	-	-	ns
SCLK rise/fall to CSN deasserted in Controller mode	Tofhm		(See Figure 10.22.1, Figure 10.22.2)	T×m -50	-	-	ns
SCLK rise/fall to CSN deasserted in Target mode	Tifhs		(See Figure 10.22.3, Figure 10.22.4)	T×n -20	-	-	ns

Note:

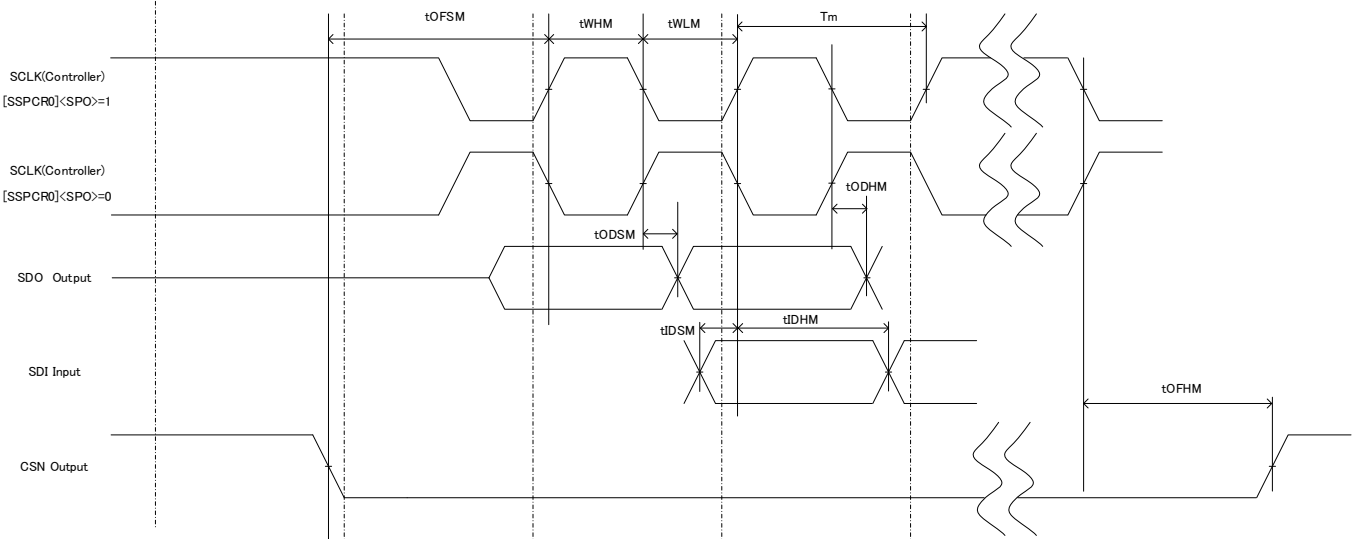
- Load capacitance, CL = 100 pF
- [GPOPSR]< GPIOPSx>=11 (6mA setting)
- The electric character of Section 39.22 is Guaranteed by design.
- T represents the SSPCLK period (e.g., 25 ns at 40 MHz).
- n represents a ratio of the SCLK period to the SSPCLK period ( $n \geq 12$ ).
- m represents the ratio of the SCLK period to the SSPCLK period ( $65024 \geq m \geq 12$ ).

Controller[SSPCR0]<SPH>=0 (Latch data on 1st edge)

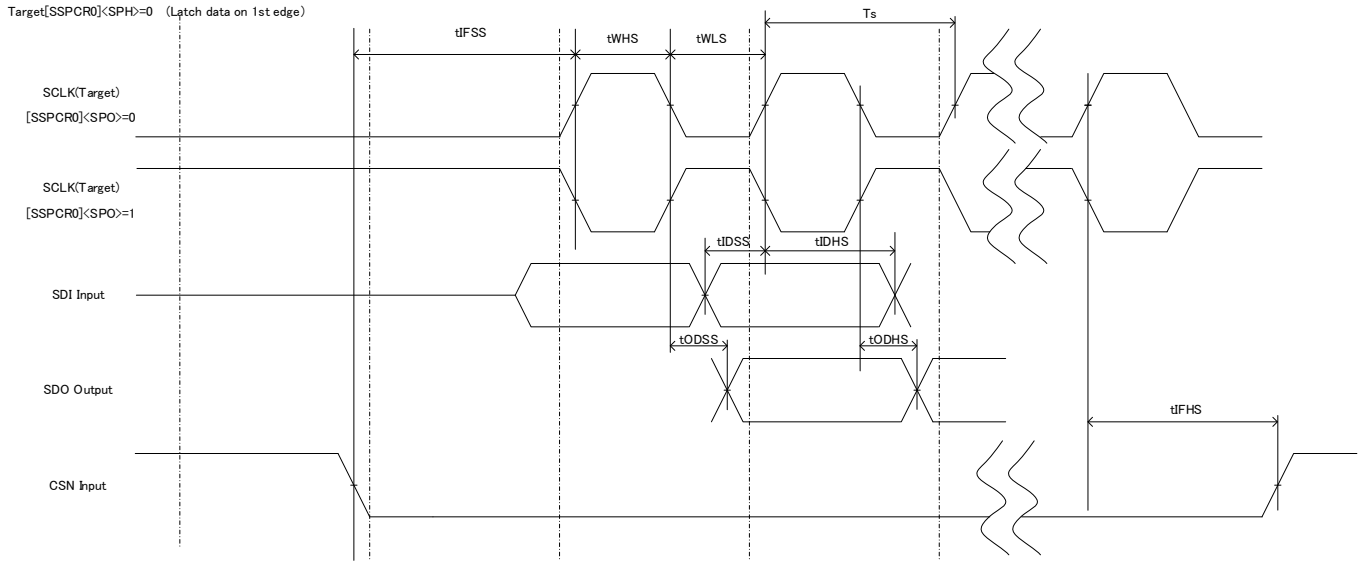


**Figure 10.22.1 SPI Communication Waveform 1**

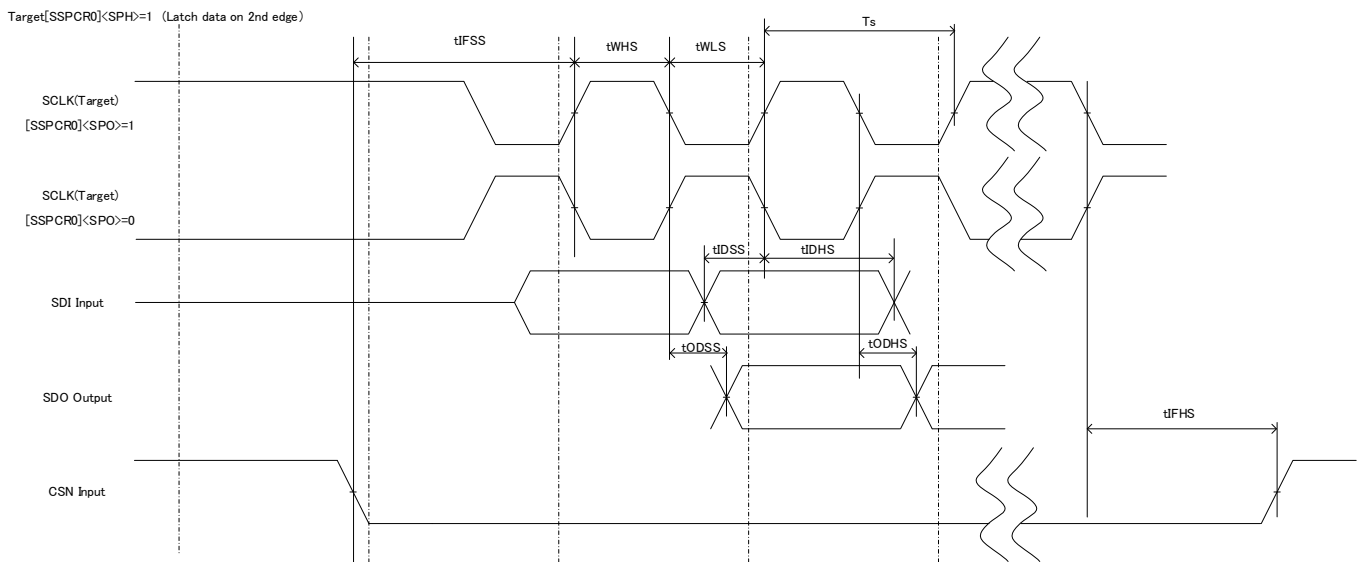
Controller[SSPCR0]<SPH>=1 (Latch data on 2nd edge)



**Figure 10.22.2 SPI Communication Waveform 2**



**Figure 10.22.3 SPI Communication Waveform 3**



**Figure 10.22.4 SPI Communication Waveform 4**

**10.23. GPIO**

**Table 10.23 Electrical Characteristics**

V<sub>BAT</sub>=6 to 18 V, V<sub>CC</sub>=4.8 to 5.2 V, V<sub>DD</sub>=1.45 to 1.55 V, and T<sub>j</sub>=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
High-level input threshold voltage	V <sub>inh</sub> *	GPIO_x	-	V <sub>CC</sub> ×0.75	-	-	V
Low-level input threshold voltage	V <sub>inl</sub> *		-	-	-	V <sub>CC</sub> ×0.25	V
Pull-up resistor	R <sub>pullup</sub>		-	30	50	100	kΩ
Pull-down resistor	R <sub>pulldown</sub>		-	30	50	100	kΩ
High-level output voltage	V <sub>outh</sub>		Load condition: [GPOPSR]<GPIOPSx>=00: -1 mA [GPOPSR]<GPIOPSx>=01: -2 mA [GPOPSR]<GPIOPSx>=10: -4 mA [GPOPSR]<GPIOPSx>=11: -6 mA	V <sub>CC</sub> ×0.8	-	-	V
Low-level output voltage	V <sub>outl</sub>		Load condition: [GPOPSR]<GPIOPSx>=00: 1 mA [GPOPSR]<GPIOPSx>=01: 2 mA [GPOPSR]<GPIOPSx>=10: 4 mA [GPOPSR]<GPIOPSx>=11: 6 mA	-	-	0.5	V

**10.24. Flash (Code Flash)**

**Table 10.24 Electrical Characteristics**

V<sub>bat</sub>=6 to 18 V, V<sub>CC</sub>=4.8 to 5.2 V, V<sub>DD</sub>=1.45 to 1.55 V, and T<sub>j</sub>=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Conditions	Values			Unit
				Min.	Typ.	Max.	
Bus frequency	-	-	-	-	-	42	MHz
Read frequency	-	-	-	-	-	10.5	MHz
Data retention time 1	-	-	T <sub>j</sub> =85°C, after 1,000 program/erase cycles	40	-	-	years
Flash capacity	-	-	-	-	64	-	KBytes
Data access size	-	-	Same as read/write operations	-	Word (32 bits)	-	-
Erase time	-	-	64KBytes B, T <sub>j</sub> = -40 to 150°C	-	80	-	ms
Program time	-	-	T <sub>j</sub> = -40 to 150°C	-	2	-	s
Erase block size	-	-	-	-	8	-	KBytes
Erase block time	-	-	1 block (8KBytes), T <sub>j</sub> = -40 to 150°C	-	6.8	-	ms
Program block size	-	-	-	-	128	-	Bytes
Program block time	-	-	128Bytes, T <sub>j</sub> = -40 to 150°C	-	2.4	-	ms

Note: It is necessary to set the number of wait cycles for read access according to the bus frequency.

## 10.25. Other

Table 10.25 Electrical Characteristics

V<sub>BAT</sub>=6 to 18 V, V<sub>CC</sub>=4.8 to 5.2 V, V<sub>DD</sub>=1.45 to 1.55 V, and T<sub>j</sub>=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
High-level input threshold voltage	V <sub>md*inh</sub>	MD0 MD1	-	V <sub>CC</sub> ×0.75	-	-	V
Low-level input threshold voltage	V <sub>md*inl</sub>		-	-	-	V <sub>CC</sub> ×0.25	V
Pull-down resistor	R <sub>md*pd</sub>		-	30	50	100	kΩ

## 10.26. Thermal resistance

Table 10.26 Electrical Characteristics

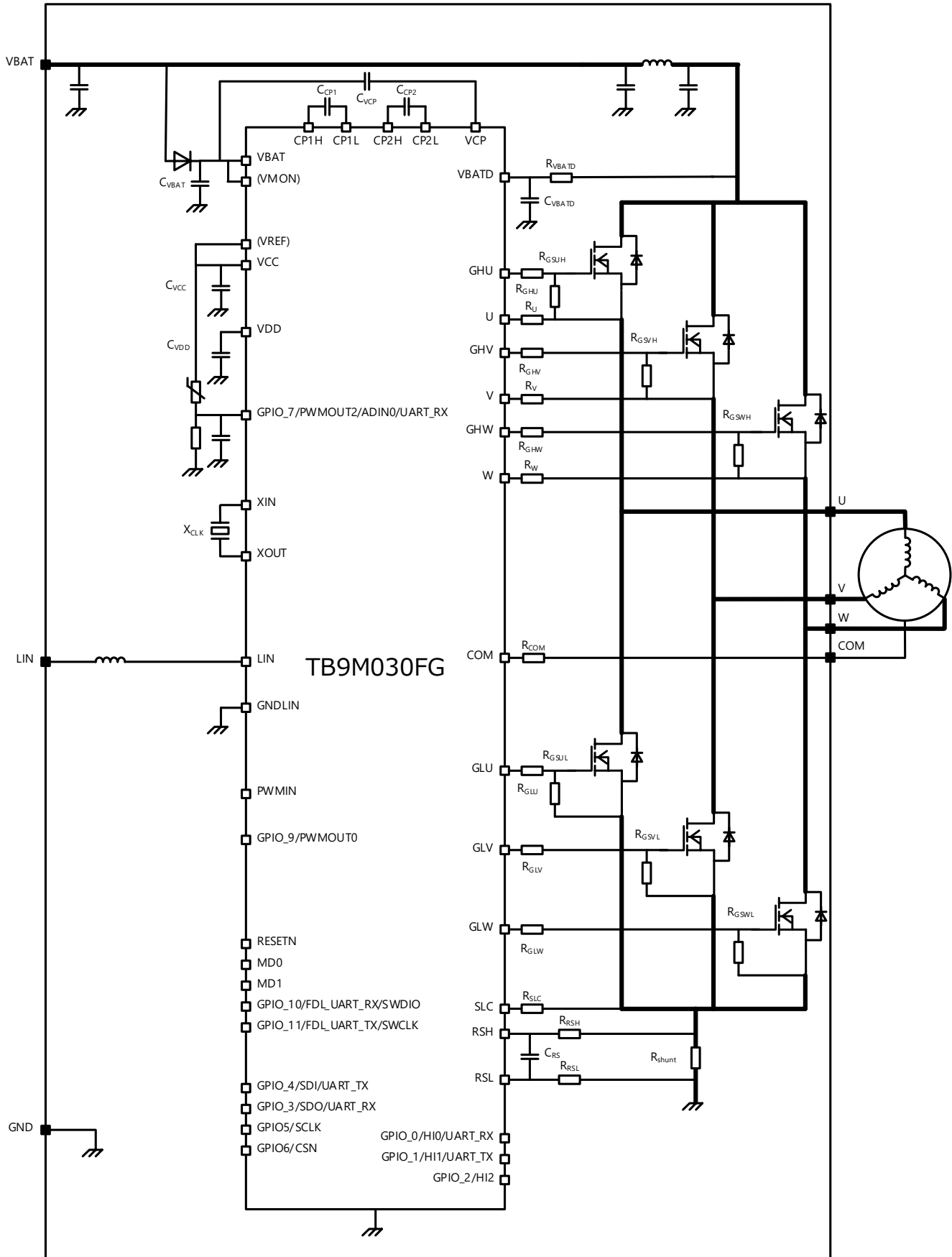
V<sub>BAT</sub>=6 to 18 V, V<sub>CC</sub>=4.8 to 5.2 V, V<sub>DD</sub>=1.45 to 1.55 V, and T<sub>j</sub>=-40 to 175°C unless otherwise noted

Characteristic	Symbol	Pin	Test Condition	Min	Typ.	Max	Unit
Junction to Ambient	θ <sub>JA</sub>	-	Thermal resistance between Junction and Ambient	-	30	-	°C/W
Junction to Package Top	ψ <sub>JT</sub>		Thermal resistance between Junction and Package Top	-	0.21	-	°C/W

Note:

- This value is the result of thermal simulation and is a guaranteed design value. Shipping inspection will not be conducted.
- The board conditions are as follows.
  - Board Size : 114.3mm × 76.2mm × 1.6mm (JEDEC Board : JESD51-7)
  - Number of layers : Multi-Layer (Cu 4 layer)
  - Cu layer thickness : 35μm (2/3 layer), 70μm (1/4 layer)
  - Cu layer area : 74.2 × 74.2 mm<sup>2</sup>
  - Number of Cu VIA : 16 (4mm × 4mm)
  - Ambient temperature: 25°C (No wind)

**11. Application Circuit Example**



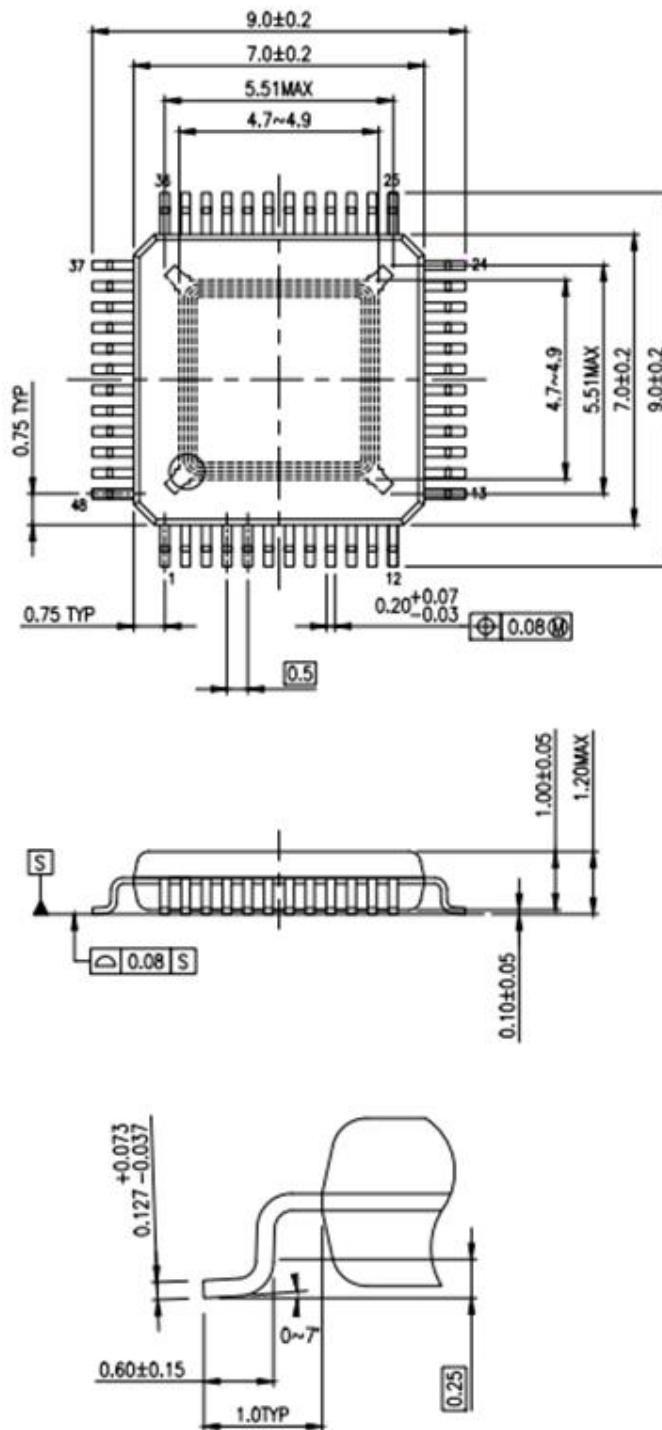
**Figure 11.1 Application Circuit Example**

**12. Package Information**

**12.1. Package Dimensions**

P-HTQFP48-0707-0.50-002

"Unit:mm"



Weight: 0.14 g (typ.)

**Figure 12.1 Package Dimensions**

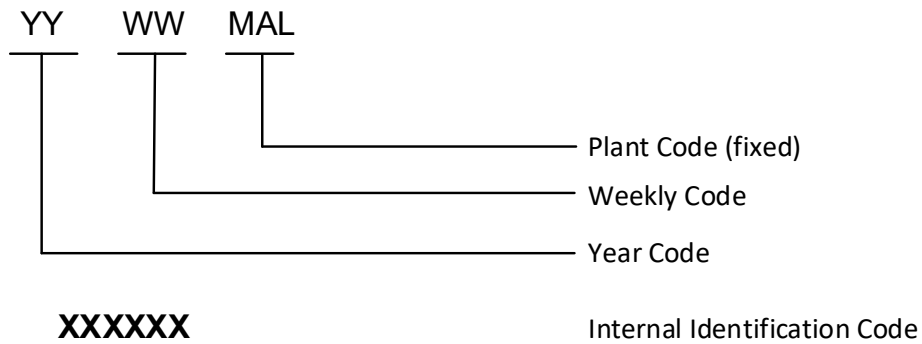
**12.2. Marking**

Product name : TB9M030FG



**Figure 12.2 Marking**

Example: Lot code breakdown



## **13. IC Usage Considerations**

### **13.1. Notes on Handling of ICs**

- (1) The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment.
- (2) Use an appropriate power supply fuse to ensure that a large current does not continuously flow in case of over current and/or IC failure.

### **13.2. Points to Remember on Handling of ICs**

- (1) **Over current Protection Circuit**  
Over current protection circuits (referred to as current limiter circuits) do not necessarily protect ICs under all circumstances. If the Over current protection circuits operate against the over current, clear the over current status immediately.
- (2) **Thermal Shutdown Circuit**  
Thermal shutdown circuits do not necessarily protect ICs under all circumstances. If the thermal shutdown circuits operate against the over temperature clears the heat generation status immediately.

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