TOSHIBA BiCD Integrated Circuit Silicon Monolithic

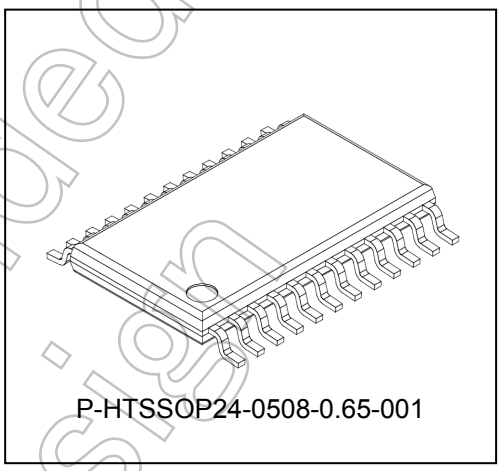
# TB62210FNG

BiCD Constant-Current Two-Phase Bipolar Stepping Motor Driver IC

The TB62210FNG is a two-phase bipolar stepping motor driver using a PWM chopper.

Fabricated with the BiCD process, the TB62210FNG is rated at 40 V/1.0 A.

The on-chip voltage regulator allows control of a stepping motor with a single $V_M$ power supply.



P-HTSSOP24-0508-0.65-001

Wight: 0.1 g (typ.)

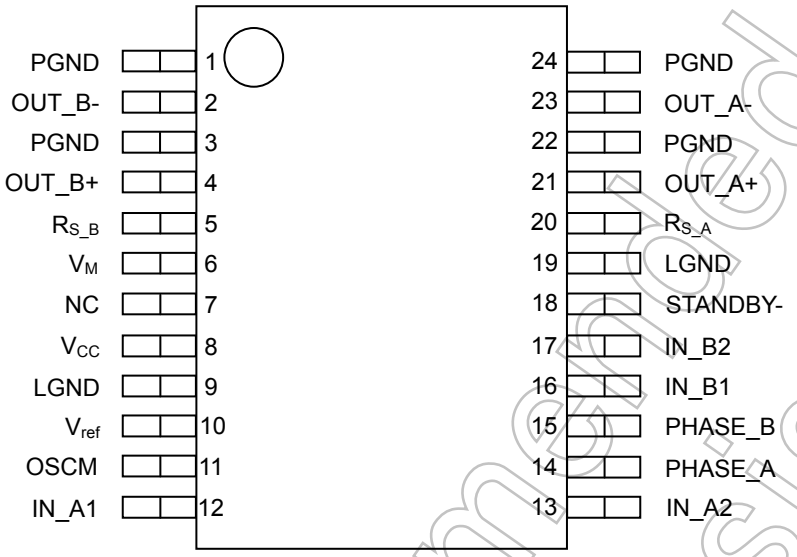
## Features

- Bipolar stepping motor driver
- PWM controlled constant-current drive
- Phase in control
- Allows two-phase, 1-2-phase and W1-2 phase excitations.
- BiCD process: Uses DMOS FETs as output power transistors.
- High voltage and current : 40 V/1.0 A (absolute maximum ratings)
- Thermal shutdown (TSD), Over current detection (ISD), and Power-on resets (PORs)
- Package: P-HTSSOP24-0508-0.65-001

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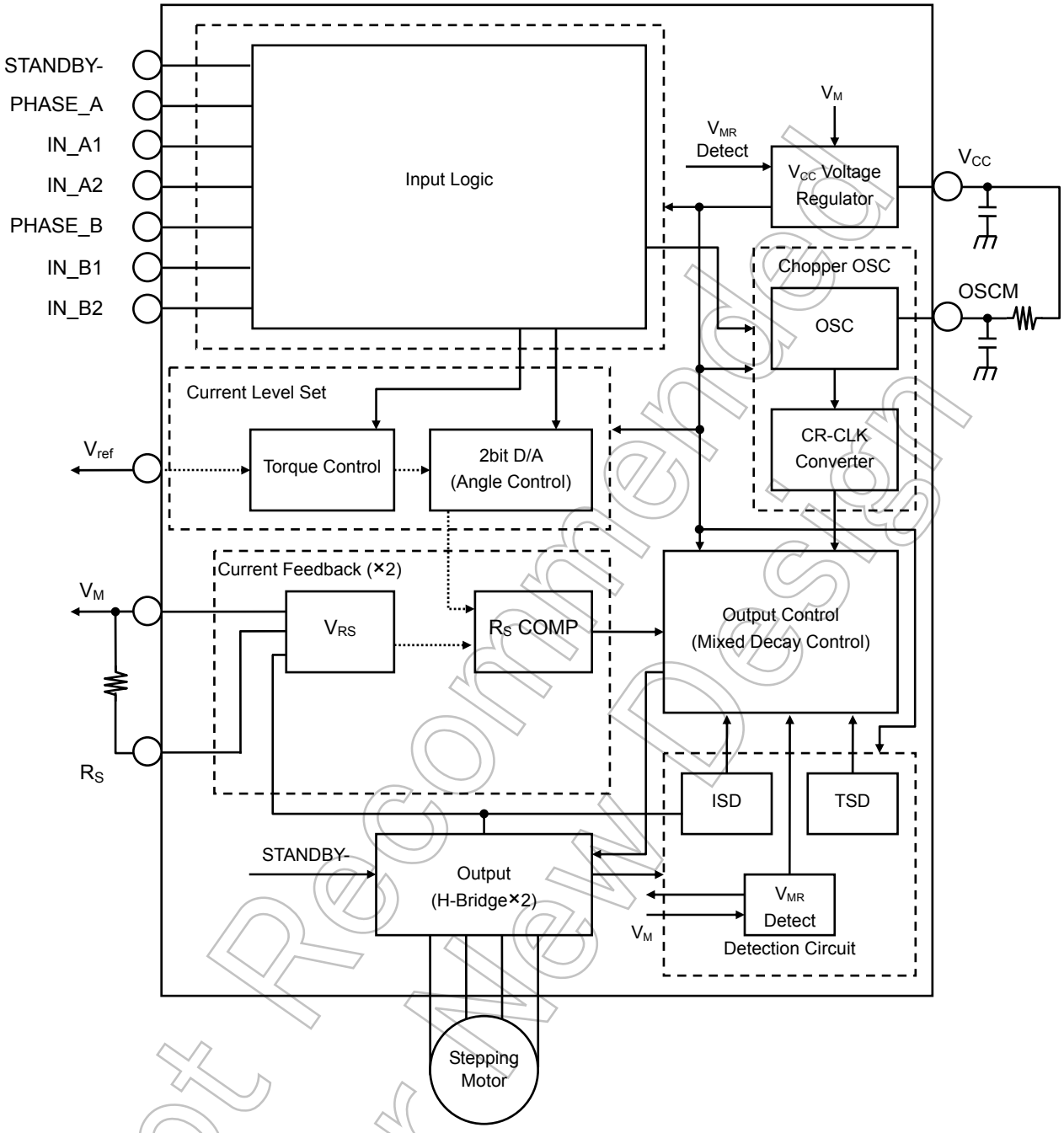
## Pin Assignment

(Top View)




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**Block Diagram**




In the block diagram, part of the functional blocks or constants may be omitted or simplified for explanatory purposes.

Note: All the grounding wires of the TB62210FNG must run on the solder mask on the PCB and be externally terminated at only one point. Also, a grounding method should be considered for efficient heat dissipation.Careful attention should be paid to the layout of the output, $V_M$ and GND traces, to avoid short-circuits across output pins or to the power supply or ground. If such a short-circuit occurs, the TB62210FNG may be permanently damaged.

Also, utmost care should be taken for pattern designing and implementation of the TB62210FNG since it has the power supply pins ($V_M$, $R_S$, OUT, GND) particularly a large current can run through. If these pins are wired incorrectly, an operation error or even worse a destruction of the TB62210FNG may occur.

The logic input pins must be correctly wired, too; otherwise, the TB62210FNG may be damaged due to a current larger than the specified current running through the IC.

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## Pin Function

This table is a function explanation from the terminal number 1 to 24.

| Pin No. | Pin Name | Function |
|---|---|---|
| 1 | PGND | Motor power ground |
| 2 | OUT_B- | B-phase negative driver output |
| 3 | PGND | Motor power ground |
| 4 | OUT_B+ | B-phase positive driver output |
| 5 | $R_{S_B}$ | The sink current sensing of B-phase motor coil |
| 6 | $V_M$ | Motor power supply monitor |
| 7 | NC | No-connect |
| 8 | $V_{CC}$ | Smoothing filter for logic power supply |
| 9 | LGND | Logic ground |
| 10 | $V_{ref}$ | Tunes the current level for motor drive |
| 11 | OSCM | Oscillator pin for PWM choppers |
| 12 | IN_A1 | A-phase excitation control input |
| 13 | IN_A2 | A-phase excitation control input |
| 14 | PHASE_A | Current direction signal input for A-phase |
| 15 | PHASE_B | Current direction signal input for B-phase |
| 16 | IN_B1 | B-phase excitation control input |
| 17 | IN_B2 | B-phase excitation control input |
| 18 | STANDBY- | All-function-initializing and Low power dissipation mode |
| 19 | LGND | Logic ground |
| 20 | $R_{S_A}$ | The sink current sensing of A-phase motor coil |
| 21 | OUT_A+ | A-phase positive driver output |
| 22 | PGND | Motor power ground |
| 23 | OUT_A- | A-phase negative driver output |
| 24 | PGND | Motor power ground |

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## Output Function Table

### Operation explanation

| Input | | | Output | | |
|---|---|---|---|---|---|
| PHASE_A<br>PHASE_B | IN_A1<br>IN_B1 | IN_A2<br>IN_B2 | OUT_A+<br>OUT_B+ | OUT_A-<br>OUT_B- | $I_{OUT}$ |
| H | H | H | H | L | 100% |
| | H | L | H | L | 71% |
| | L | H | H | L | 38% |
| | L | L | Outputs disabled | Outputs disabled | 0% |
| L | H | H | L | H | -100% |
| | H | L | L | H | -71% |
| | L | H | L | H | -38% |
| | L | L | Outputs disabled | Outputs disabled | 0% |

$I_{OUT}$: The current which flows OUT_A+ (OUT_B+) to OUT_A-(OUT_B-) is defined plus current. The current which flows OUT_A-(OUT_B-) to OUT_A+ (OUT_B+) is defined minus current.

## Other Functions

| Pin Name | H | L | Notes |
|---|---|---|---|
| IN_A1<br>IN_A2<br>IN_B1<br>IN_B2 | Outputs enabled | Outputs disabled | When IN_X is disserted Low (where a letter X that indicates a phase), its outputs assume the high-impedance state, regardless of the state of that phase. |
| PHASE_A<br>PHASE_B | OUT_A+(OUT_B+) : H | OUT_A-(OUT_B-) : H | When PHASE is High, a current normally flows from OUT_A+ (OUT_B+) to OUT_A-(OUT_B-). |
| STANDBY- | Normal operation mode | Standby mode | When STANDBY- is Low, both the oscillator and output drivers are disabled. The TB62210FNG can not drive a motor. |

Input signals to IN_X and PHASE_X after the voltage range of the motor being used is attained.

(*X: A1, A2, B1. B2)

## Protection Features

(1) Thermal shutdown (TSD)
The thermal shutdown circuit turns off all the outputs when the junction temperature ($T_j$) exceeds 150°C (typ.).
The outputs retain the current states.
The TB62210FNG exits TSD mode and resume normal operation when the TB62210FNG is rebooted or the STANDBY- pin is changed from High to Low and then to High.

(2) Power-ON-resets (PORs) for $V_{MR}$ and $V_{CCR}$ ($V_M$ and $V_{CC}$ voltage monitor)
The outputs are forced off until $V_M$ and $V_{CC}$ reach the rated voltages.

(3) Over current detection (ISD)
Each phase has an over current detection circuit, which turns off the corresponding outputs when the output current exceeds the shutdown trip threshold (above the maximum current rating).
The TB62210FNG exits ISD mode and resume normal operation when the STANDBY- pin is changed from High to Low and then to High.
This circuit provides protection against a short-circuit by temporarily disabling the device. Important notes on this feature will be provided later.

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## Cautions on Over current Detection (ISD) and Thermal Shutdown (TSD)

• The ISD and TSD circuits are only intended to provide temporary protection against irregular conditions such as an output short-circuit; they do not necessarily guarantee the complete IC safety.

• If the device is used beyond the specified operating ranges, these circuits may not operate properly: then the device may be damaged due to an output short-circuit.

• The ISD circuit is only intended to provide a temporary protection against an output short-circuit. If such a condition persists for a long time, the device may be damaged due to overstress. Over current conditions must be removed immediately by external hardware.

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## Absolute Maximum Ratings ($T_a$ = 25°C)

| Characteristics | Symbol | Rating | Unit |
|---|---|---|---|
| Motor power supply | $V_M$ | 40 | V |
| Motor output voltage | $V_{OUT}$ | 40 | V |
| Motor output current (Note 1) | $I_{OUT}$ | 1.0 | A per phase |
| Logic power supply | $V_{CC}$ | 6.0 | V |
| Logic input voltage | $V_{IN}$ | 6.0 | V |
| $V_{ref}$ reference voltage | $V_{ref}$ | 5.0 | V |
| Power dissipation (Note 2) | $P_D$ | 3.125 | W |
| Operating temperature | $T_{opr}$ | -20 to 85 | °C |
| Storage temperature | $T_{stg}$ | -55 to 150 | °C |
| Junction temperature | $T_{j(MAX)}$ | 150 | °C |

Note 1: Perform thermal calculations for the maximum current value under normal conditions. Use the IC at 0.6 A or less per phase. The current value maybe controlled according to the ambient temperature or board conditions.

Note 2: Mounts on the substrate ($T_a$ = 25°C)
If $T_a$ is over 25°C, derating is required at 25mW/°C.
$T_a$: Ambient temperature
$T_{opr}$: Ambient temperature while the TB62210FNG is active.
$T_j$: Junction temperature while the TB62210FNG is active. The maximum junction temperature is limited by the thermal shutdown (TSD) circuitry.
Because the maximum value of $T_j$ is 120°C, recommended maximum current usage is below 120°C.

### Absolute Maximum Ratings

The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment. Do not exceed any of these ratings.
Exceeding the rating(s) may cause the device breakdown, damage or deterioration, and may result injury by explosion or combustion.
The value of even one parameter of the absolute maximum ratings should not be exceeded under any circumstances. The TB62210FNG does not have overvoltage protection. Therefore, the device is damaged if a voltage exceeding its rated maximum is applied.
All voltage ratings including supply voltages must always be followed. The other notes and considerations described later should also be referred to.

## Operating Ranges ($T_a$ = 0 to 85°C)

| Characteristics | Symbol | Min | Typ. | Max | Unit | Remarks |
|---|---|---|---|---|---|---|
| Motor power supply | $V_M$ | 10.0 | 24.0 | 38.0 | V | - |
| Motor output current | $I_{OUT}$ | - | 0.6 | 1.0 | A | Per phase (Note 1) |
| Logic input voltage | $V_{IN(H)}$ | 2.0 | - | 5.5 | V | Logic high level |
| | $V_{IN(L)}$ | -0.4 | - | 1.0 | V | Logic low level |
| Chopper frequency | $f_{chop}$ | 40 | - | 150 | kHz | - |
| $V_{ref}$ reference voltage | $V_{ref}$ | GND | - | 3.6 | V | - |
| Voltage across the current-sensing resistor pins | $V_{RS}$ | 0.0 | ±1.0 | ±1.5 | V | Referenced to the $V_M$ pin (Note 2) |

Note 1: The actual maximum current may be limited by the operating environment (operating conditions such as excitation mode or operating duration, or by the surrounding temperature or board heat dissipation). Determine a realistic maximum current by calculating the heat generated under the operating environment.

Note 2: The maximum $V_{RS}$ voltage should not exceed the maximum rated voltage.

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## Electrical Characteristics 1 ($T_a$ = 25°C, $V_M$ = 24 V, unless otherwise specified)

| Characteristics | | Symbol | Test Condition | Min | Typ. | Max | Unit |
|---|---|---|---|---|---|---|---|
| Digital input voltage | High | $V_{IN(H)}$ | Digital input pins (Note) | 2 | 3.3 | 5.5 | V |
| | Low | $V_{IN(L)}$ | Digital input pins (Note) | -0.4 | - | 1.0 | V |
| Input hysteresis voltage | | $V_{IN(HIS)}$ | Digital input pins (Note) | 100 | 200 | 300 | mV |
| Digital input current | High | $I_{IN(H)}$ | $V_{IN}$ = 5 V at the digital input pins under test | 35 | 50 | 75 | μA |
| | Low | $I_{IN(L)}$ | $V_{IN}$ = 0 V at the digital input pins under test | - | - | 1 | μA |
| Current consumption | | $I_{M1}$ | Outputs open, STANDBY- = Low | - | 2 | 3 | mA |
| | | $I_{M2}$ | Outputs open, STANDBY- = High | - | 3.5 | 5 | mA |
| | | $I_{M3}$ | Outputs open (two-phase excitation) | - | 5 | 7 | mA |
| Output leakage current | High-side | $I_{OH}$ | $V_{RS}$ = $V_M$ = 40 V, $V_{OUT}$ = 0 V | - | - | 2 | μA |
| | Low-side | $I_{OL}$ | $V_{RS}$ = $V_M$ = $V_{OUT}$ = 40 V | 2 | - | - | μA |
| Chanel-to-channel current differential | | $\Delta I_{OUT1}$ | Channel-to-channel error | -5 | 0 | 5 | % |
| Output current error relative to the predetermined value | | $\Delta I_{OUT2}$ | $I_{OUT}$ = 1.0 A | -7 | 0 | 7 | % |
| $R_S$ pin current | | $I_{RS}$ | $V_{RS}$ = $V_M$ = 24 V | 0 | - | 10 | μA |
| Drain-source ON-resistance of the output transistors (upper and lower sum) | | $R_{ON(D-S)}$ | $I_{OUT}$ = 1.0A, $T_j$ = 25°C | - | 1.2 | 1.5 | Ω |

Note: $V_{IN\ (L \rightarrow H)}$ is defined as the $V_{IN}$ voltage that causes the outputs to change when a pin under test is gradually raised from 0 V. $V_{IN\ (H \rightarrow L)}$ is defined as the $V_{IN}$ voltage that causes the outputs to change when the pin is then gradually lowered.
The difference between $V_{IN\ (L \rightarrow H)}$ and $V_{IN\ (H \rightarrow L)}$ is defined as the input hysteresis.

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## Electrical Characteristics 2 ($T_a$ = 25°C, $V_M$ = 24 V, unless otherwise specified)

| Characteristics | Symbol | Test Condition | Min | Typ. | Max | Unit |
|---|---|---|---|---|---|---|
| $V_{ref}$ input current | $I_{ref}$ | $V_{ref}$ = 3.0 V | - | 0 | 1 | μA |
| $V_{ref}$ decay rate | $V_{ref}$(gain) | $V_{ref}$ = 2.0 V | 1/4.8 | 1/5.0 | 1/5.2 | - |
| TSD threshold (Note 1) | $T_j$TSD | - | 140 | 150 | 170 | °C |
| $V_M$ recovery voltage (Note 2) | $V_{MR}$ | - | 7.0 | 8.0 | 9.0 | V |
| Supply voltage for internal circuitry (Note 3) | $V_{CC}$ | $I_{CC}$ = 5.0 mA | 4.75 | 5.00 | 5.25 | V |

Note 1: Thermal shutdown (TSD) circuitry
When the junction temperature of the device has reached the threshold, the TSD circuitry is tripped, causing the internal reset circuitry to turn off the output transistors.
The TSD circuitry is tripped at a temperature between 140°C (min) and 170°C (max). Once tripped, the TSD circuitry keeps the output transistors off until STANDBY- is deasserted High or the IC is restarted. The thermal shutdown circuit is provided to turn off all the outputs when the IC is overheated. For this reason, please avoid using TSD for other purposes.

Note 2: The circuit design has been designed so that electromotive force or leak current from signal input does not occur when $V_M$ voltage is not supplied, even if the logic input signal is input. Even so, regulate logic input signals before resupply of $V_M$ voltage so that the motor does not operate when voltage is reapplied.

Note 3: If the supply voltage for internal circuitry ($V_{CC}$) is split with an external resistor and used as $V_{ref}$ input supply voltage, the accuracy of the output current setting will be at ±8% when the $V_{cc}$ output voltage accuracy and the $V_{ref}$ damping ratio accuracy are combined.

## Back-EMF

While a motor is rotating, there is a timing at which power is fed back to the power supply. At that timing, the motor current recirculates back to the power supply due to the effect of the motor back-EMF.

If the power supply does not have enough sink capability, the power supply and output pins of the device might rise above the rated voltages. The magnitude of the motor back-EMF varies with usage conditions and motor characteristics. It must be fully verified that there is no risk that the TB62210FNG or other components will be damaged or fail due to the motor back-EMF.

## IC Mounting

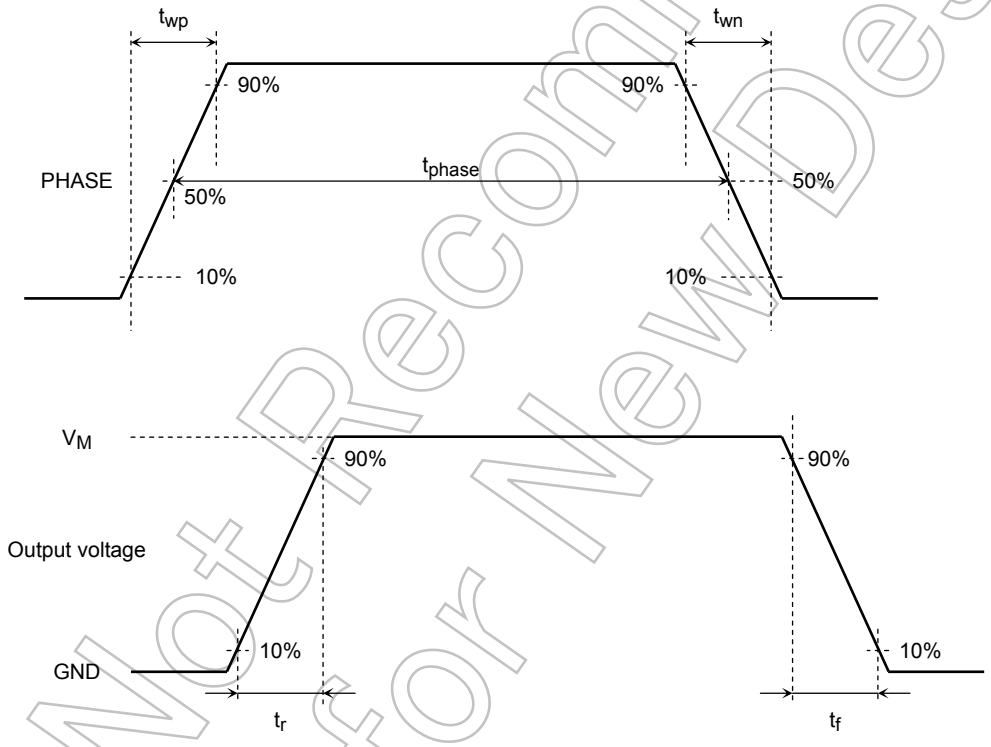
Do not insert devices in the wrong orientation or incorrectly. Otherwise, it may cause the device breakdown, damage and/or deterioration.

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## AC Electrical Characteristics ($T_a$ = 25°C, $V_M$ = 24 V, 6.8 mH/5.7 Ω)

| Characteristics | Symbol | Test Condition | Min | Typ. | Max | Unit |
|---|---|---|---|---|---|---|
| Phase frequency | $f_{PHASE}$ | $f_{CR}$ = 1600 kHz | - | - | 400 | kHz |
| Minimum phase pulse width | $t_{PHASE}$ | - | 100 | - | - | ns |
| | $t_{wp}$ | - | 50 | - | - | ns |
| | $t_{wn}$ | - | 50 | - | - | ns |
| Output transistor switching characteristics | $t_r$ | - | 150 | 200 | 250 | ns |
| | $t_f$ | - | 100 | 150 | 200 | ns |
| Blanking time for current spike prevention | $t_{BLANK}$ | $I_{OUT}$ = 1.0 A | 200 | 300 | 500 | ns |
| OSCM oscillation frequency | $f_{CR}$ | $C_{OSC}$ = 270 pF, $R_{OSC}$ = 3.6 kΩ | 1200 | 1600 | 2000 | kHz |
| Chopper frequency range | $f_{chop\ (RANGE)}$ | $V_M$ = 24 V, outputs enabled ACTIVE ($I_{OUT}$ = 1.0 A) | 40 | - | 150 | kHz |
| Predefined chopper frequency | $f_{chop}$ | Outputs enabled ACTIVE ($I_{OUT}$ = 1.0 A), $f_{CR}$ = 1600 kHz | - | 100 | - | kHz |

### Timing Charts of Output Transistors Switching




The timing charts may be simplified for explanatory purpose.

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## Calculation of the Predefined Output Current

For PWM constant-current control, the TB62210FNG uses a clock generated by the CR oscillator. The peak output current can be set via the current-sensing resistor ($R_{RS}$) and the reference voltage ($V_{ref}$), as follows:

$$I_{OUT} = (1 / 5) \times V_{ref}[V] \times (1 / R_{RS}[\Omega])$$

where, 1/5 is the $V_{ref}$ decay rate, $V_{ref\ (gain)}$. For the value of $V_{ref\ (gain)}$, see the Electrical Characteristics table.

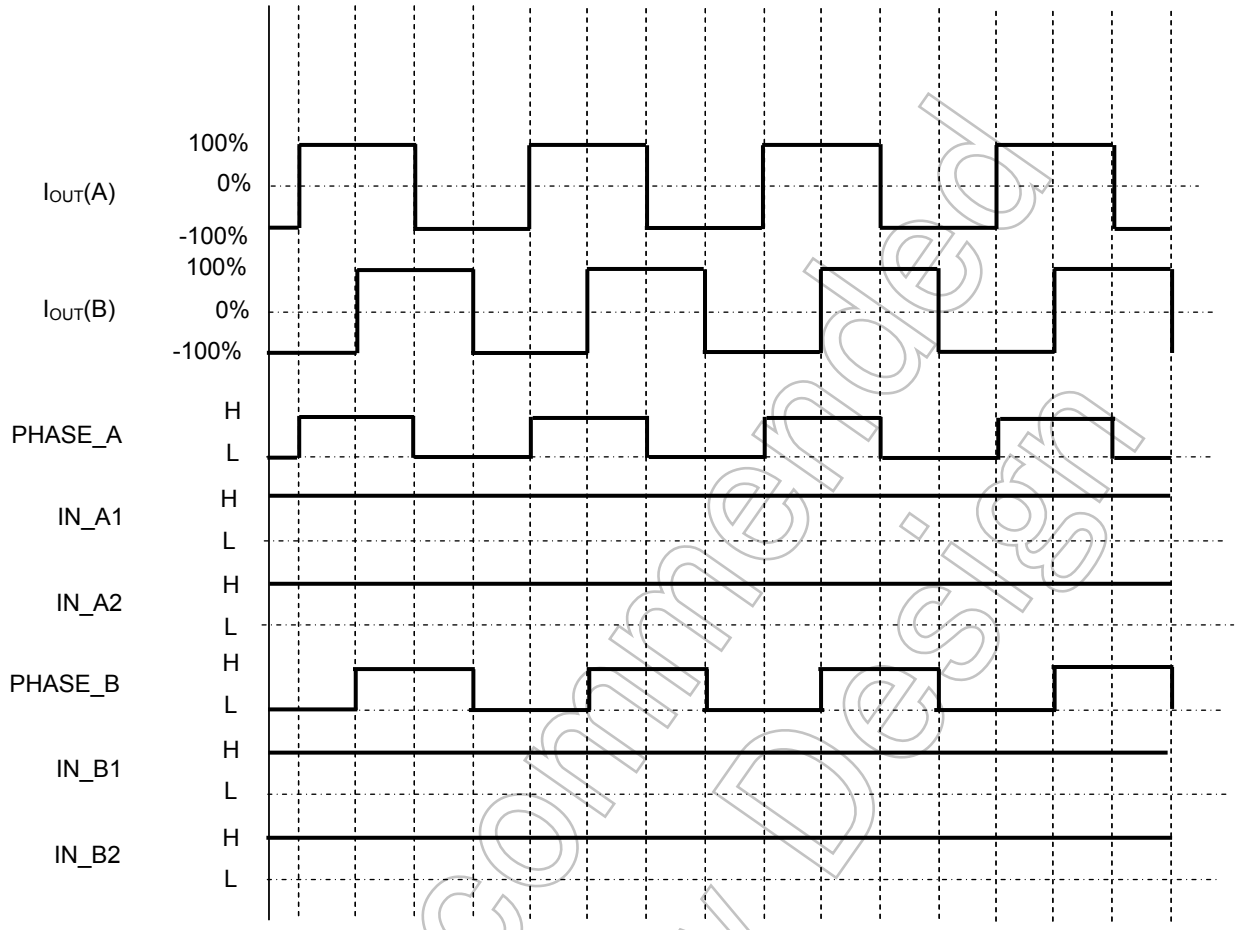
For example, when $V_{ref} = 0.88$ V, to generate an output current ($I_{OUT}$) of 0.8 A, $R_{RS}$ is calculated as:

$R_{RS} = (V_{ref} /5) / I_{OUT} = (0.88 / 5) / 0.8 = 0.22\Omega$ ($\geq$ 0.5W)

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## Phase Sequences

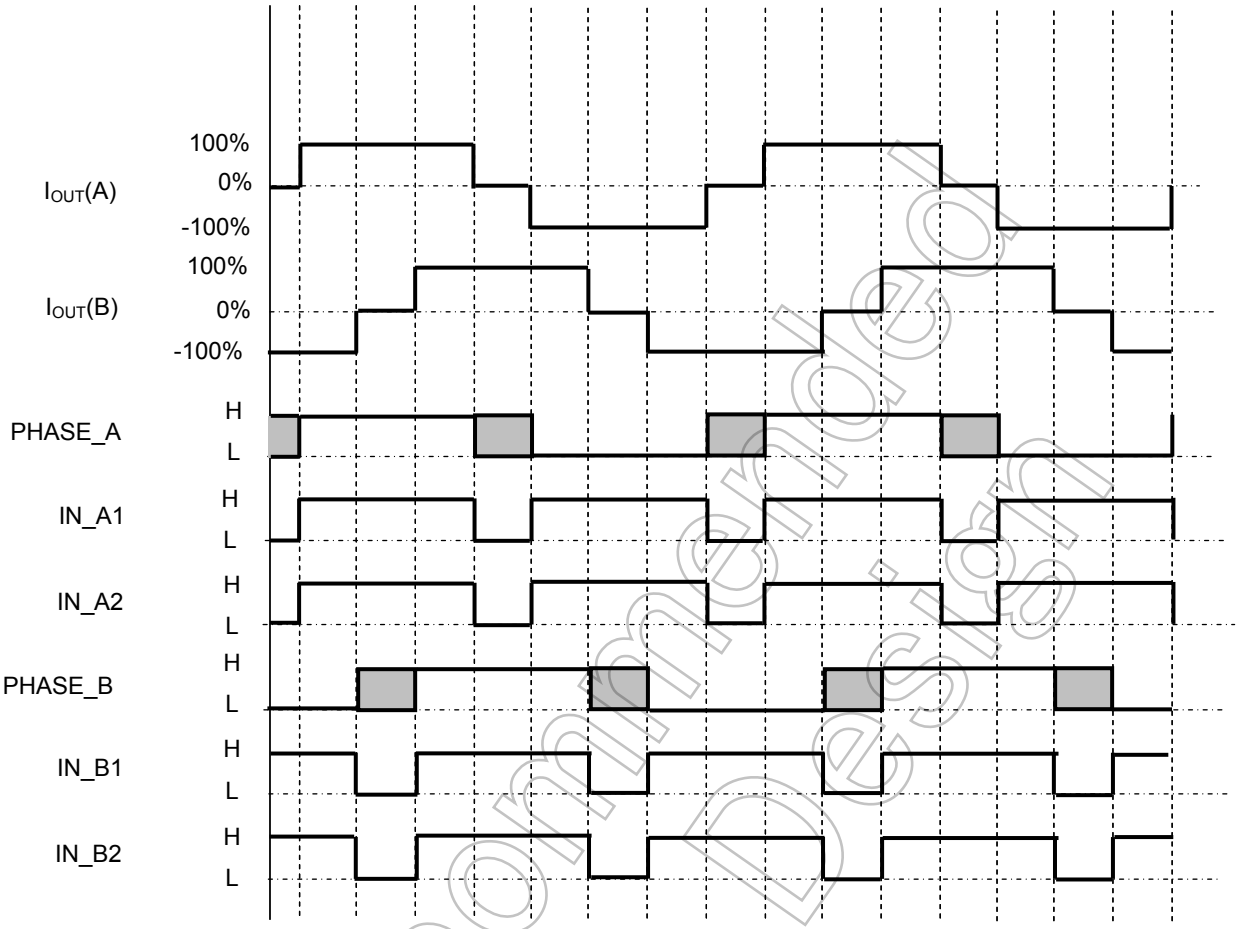
### 2-Phase Excitation Mode




Timing charts may be simplified for explanatory purposes.

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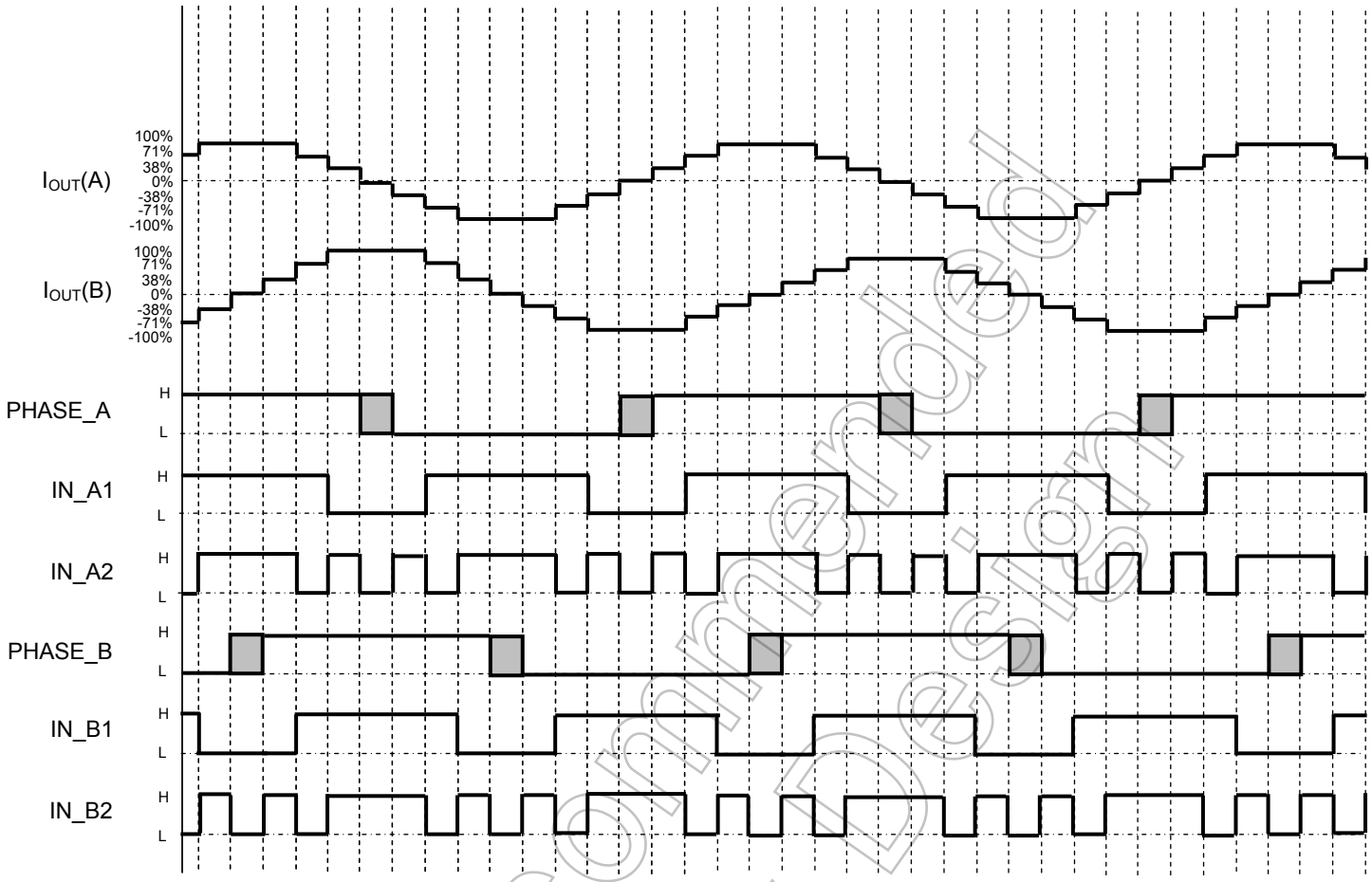
**1-2-Phase Excitation**




Timing charts may be simplified for explanatory purposes.

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**W1-2-Phase Excitation**




Timing charts may be simplified for explanatory purposes.

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## Calculation of the OSCM oscillation frequency (chopper reference frequency)

The OSCM oscillating frequency can be approximated using the following equation:

$$f_{CR} = 1 / [0.56 \times C_{OSC} \times (R_{OSC} + 500)]$$

Where:

$C_{OSC}$ = Capacitor capacity

$R_{OSC}$ = Resistance

Assigning $C_{OSC} = 270 \times 10^{-12}$ [F], $R_{OSC} = 3600$ [Ω] to get:

$f_{CR} = 1.61 \times 10^{6} = 1.6$ MHz

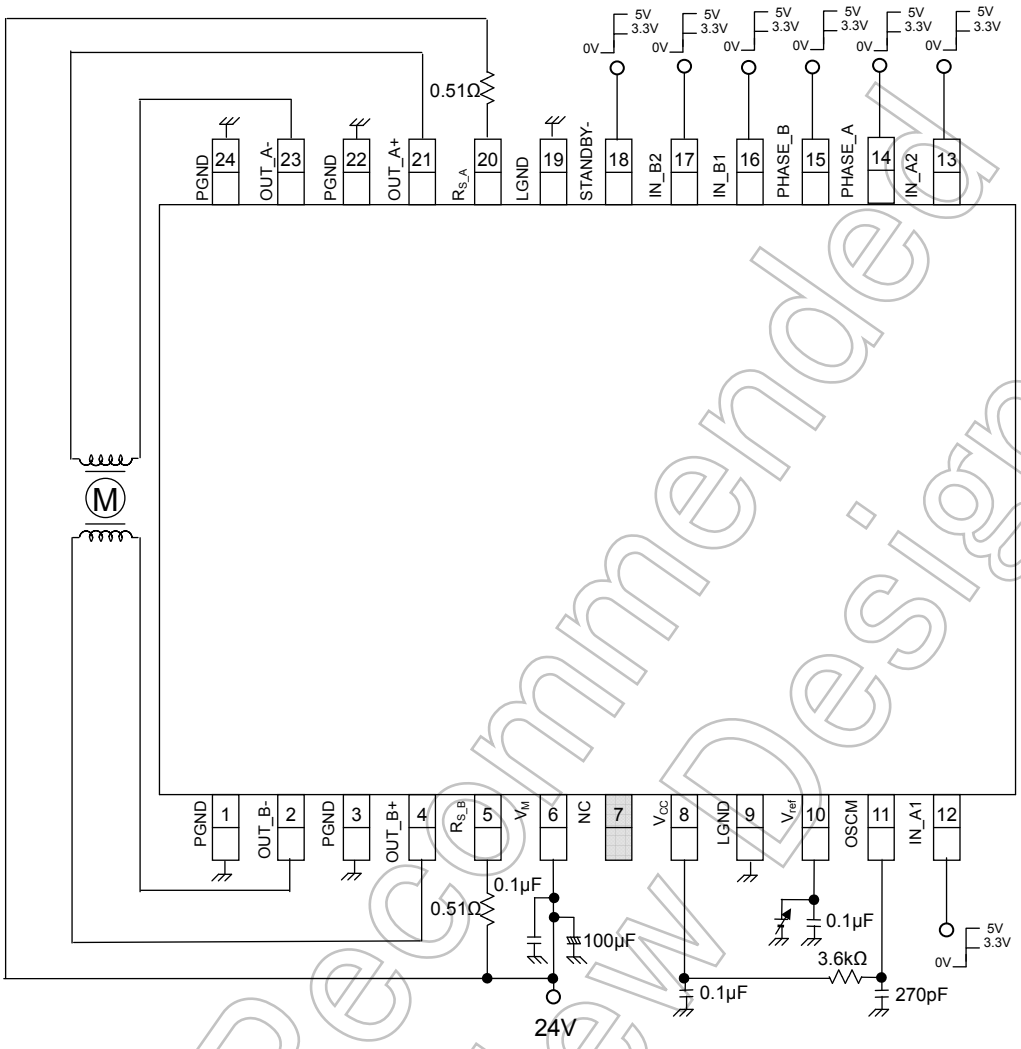
At this time, the chopping frequency fchop is calculated as follows:

$$fchop = f_{CR}/16 = 100 \text{ kHz}$$

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## Application Circuit Example

The values shown in the following figure are typical values. For input conditions, see Operating Ranges.




Note: I will recommend the addition of capacitor if necessary. The GND wiring must become one point as much as possible-earth.

The example of applied circuit is an example of the reference, and do an enough evaluation before the mass production design, please.
Moreover, it is not the one to permit the use of the industrial property.

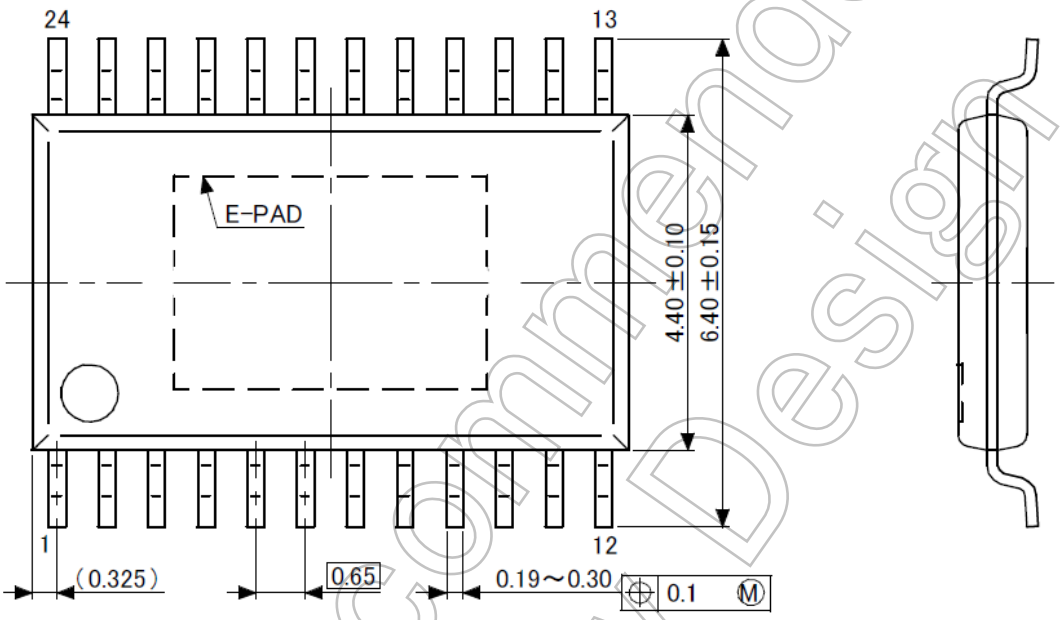
The IC may be destroyed due to short circuit between output pins, an output pin and the VDD pin, or an output pin and the GND pin. Design an output line, VM line and GND line with great care. Also a low-withstand-voltage device may be destroyed when mounted in the wrong orientation, which causes high-withstanding voltage to be applied to the device.

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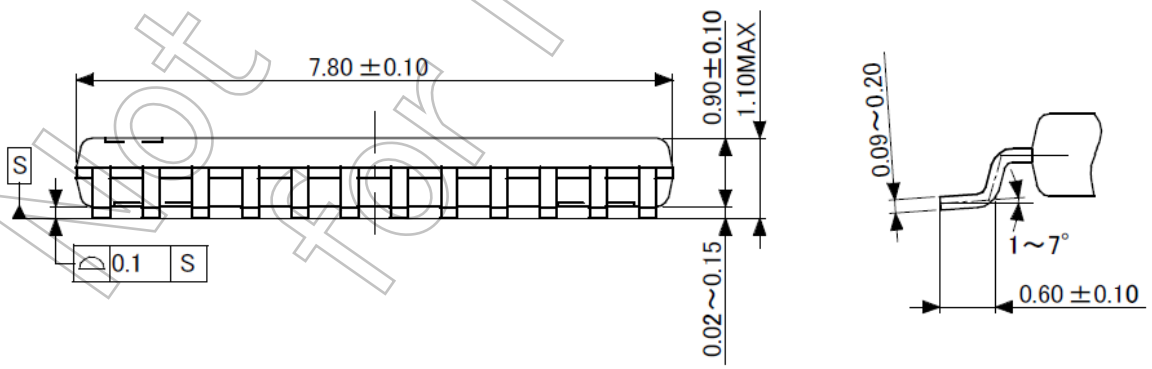
## Package Dimensions

P-HTSSOP24-0508-0.65-001

"Unit : mm"




The heat sink is located on the back side and has dimensions of 2.85 mm × 4.05 mm.




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## Notes on Contents

### Block Diagrams

Some of the functional blocks, circuits, or constants in the block diagram may be omitted or simplified for explanatory purposes.

### Timing Charts

Timing charts may be simplified for explanatory purposes.

### Application Circuits

The application circuits shown in this document are provided for reference purposes only. Thorough evaluation is required, especially at the mass production design stage.

Toshiba does not grant any license to any industrial property rights by providing these examples of application circuits.

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## IC Usage Considerations

### Notes on handling of ICs

(1) The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment. Do not exceed any of these ratings.
Exceeding the rating(s) may cause the device breakdown, damage or deterioration, and may result injury by explosion or combustion.

(2) Use an appropriate power supply fuse to ensure that a large current does not continuously flow in case of over current and/or IC failure. The IC will fully break down when used under conditions that exceed its absolute maximum ratings, when the wiring is routed improperly or when an abnormal pulse noise occurs from the wiring or load, causing a large current to continuously flow and the breakdown can lead smoke or ignition. To minimize the effects of the flow of a large current in case of breakdown, appropriate settings, such as fuse capacity, fusing time and insertion circuit location, are required.

(3) If your design includes an inductive load such as a motor coil, incorporate a protection circuit into the design to prevent device malfunction or breakdown caused by the current resulting from the inrush current at power ON or the negative current resulting from the back electromotive force at power OFF. IC breakdown may cause injury, smoke or ignition.
Use a stable power supply with ICs with built-in protection functions. If the power supply is unstable, the protection function may not operate, causing IC breakdown. IC breakdown may cause injury, smoke or ignition.

(4) Do not insert devices in the wrong orientation or incorrectly.
Make sure that the positive and negative terminals of power supplies are connected properly.
Otherwise, the current or power consumption may exceed the absolute maximum rating, and exceeding the rating(s) may cause the device breakdown, damage or deterioration, and may result injury by explosion or combustion.
In addition, do not use any device that is applied the current with inserting in the wrong orientation or incorrectly even just one time.

(5) Carefully select external components (such as inputs and negative feedback capacitors) and load components (such as speakers), for example, power amp and regulator.
If there is a large amount of leakage current such as input or negative feedback condenser, the IC output DC voltage will increase. If this output voltage is connected to a speaker with low input withstand voltage, over current or IC failure can cause smoke or ignition. (The over current can cause smoke or ignition from the IC itself.) In particular, please pay attention when using a Bridge Tied Load (BTL) connection type IC that inputs output DC voltage to a speaker directly.

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## Points to remember on handling of ICs

(1) Over current Detection Circuit

Over current detection circuits (referred to as current limiter circuits) do not necessarily protect ICs under all circumstances. If the Over current detection circuits operate against the over current, clear the over current status immediately.

Depending on the method of use and usage conditions, such as exceeding absolute maximum ratings can cause the over current detection circuit to not operate properly or IC breakdown before operation. In addition, depending on the method of use and usage conditions, if over current continues to flow for a long time after operation, the IC may generate heat resulting in breakdown.

(2) Thermal Shutdown Circuit

Thermal shutdown circuits do not necessarily protect ICs under all circumstances. If the thermal shutdown circuits operate against the over temperature, clear the heat generation status immediately.

Depending on the method of use and usage conditions, such as exceeding absolute maximum ratings can cause the thermal shutdown circuit to not operate properly or IC breakdown before operation.

(3) Heat Radiation Design

In using an IC with large current flow such as power amp, regulator or driver, please design the device so that heat is appropriately radiated, not to exceed the specified junction temperature (TJ) at any time and condition. These ICs generate heat even during normal use. An inadequate IC heat radiation design can lead to decrease in IC life, deterioration of IC characteristics or IC breakdown. In addition, please design the device taking into considerate the effect of IC heat radiation with peripheral components.

(4) Back-EMF

When a motor rotates in the reverse direction, stops or slows down abruptly, a current flow back to the motor's power supply due to the effect of back-EMF. If the current sink capability of the power supply is small, the device's motor power supply and output pins might be exposed to conditions beyond maximum ratings. To avoid this problem, take the effect of back-EMF into consideration in system design.

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